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# High Performance RF in Wireless Infrastructures

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# **Content**

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- **Wireless infrastructures**
  - State-of-the-art
  - Need for (further) integration
- **(Re) Quest for High Performance RF**
  - Technology
  - Design
- **Concluding remarks**

# Wireless communication

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cellular base station



cellular handheld



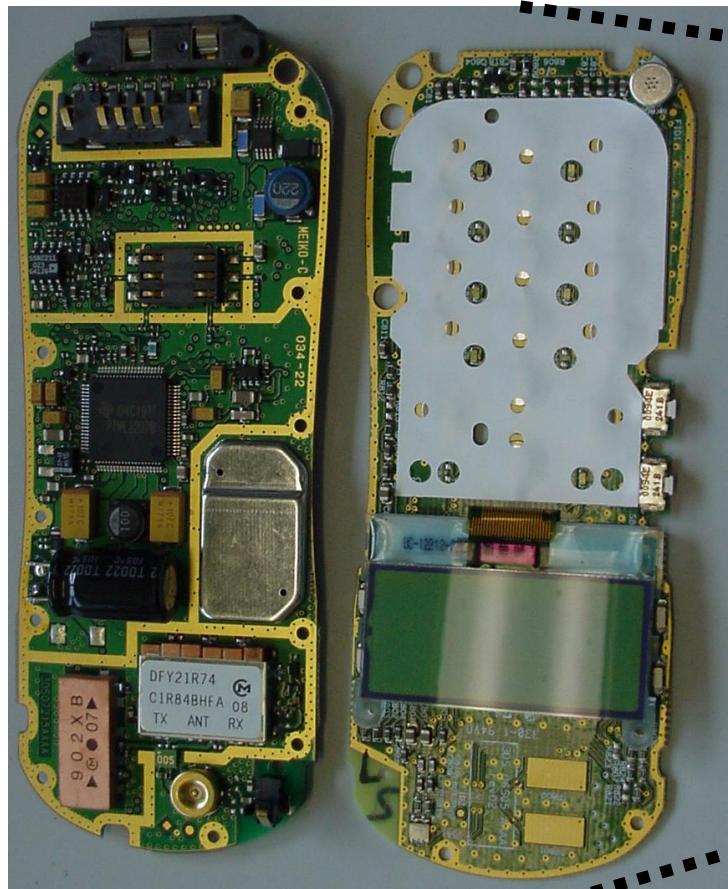
Ka/Ku band satellite



Wireless infrastructures

# Trend in cellular mobile handset

1995



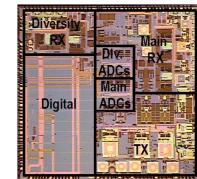
2000



2007



2009



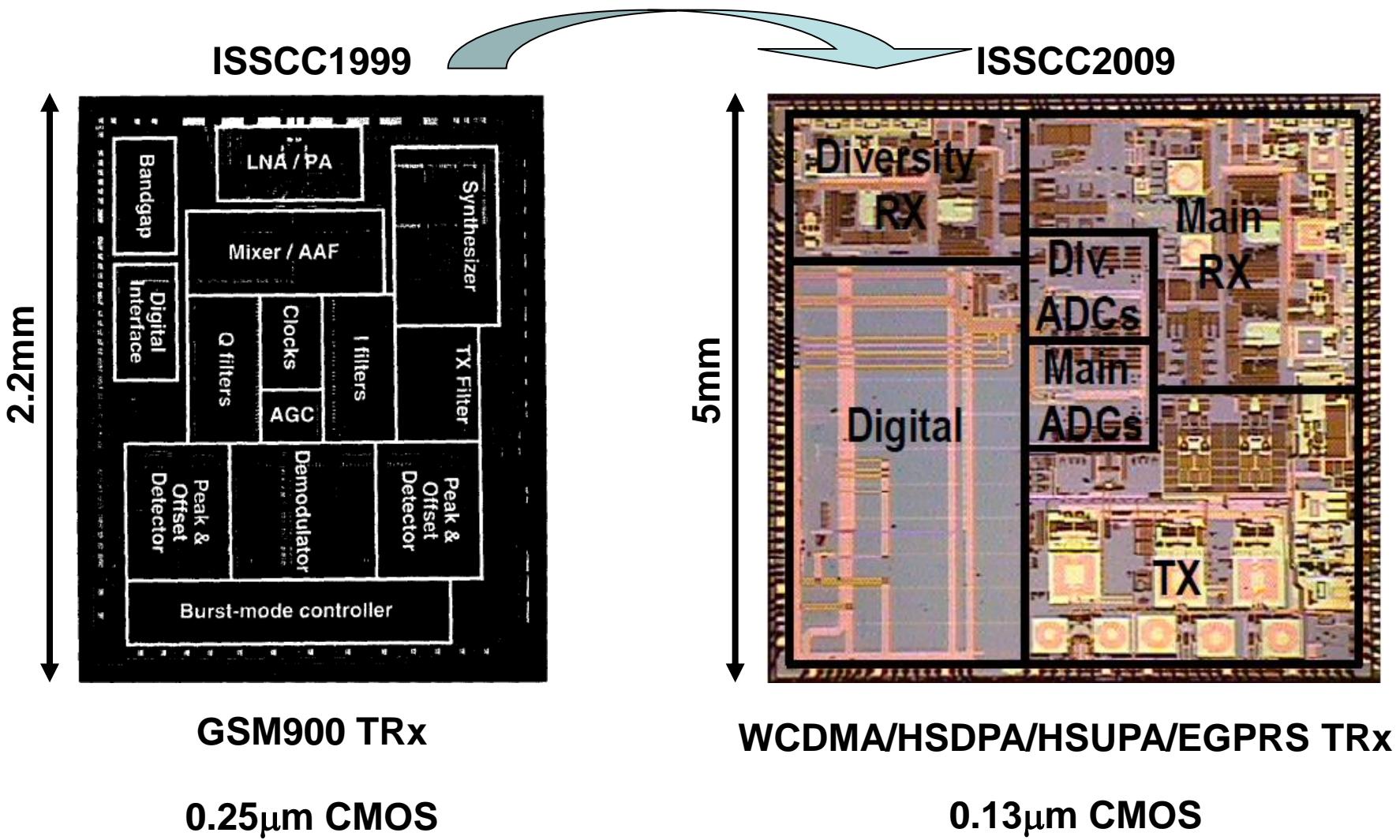
GSM900

GSM900/1800/1900

Quad EDGE  
Triple WCDMA

3.9G

# Thanks to impressive integration



# **State-of-the-art cellular handheld**

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- **Typical consumer market product**
  - Cost is key
  - Integration of functionality in Si-based technology
  - CMOS scaling lead to RF capability of devices
  - SoC solutions became possible

# Let's look at cellular base station

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Tower + cabinets

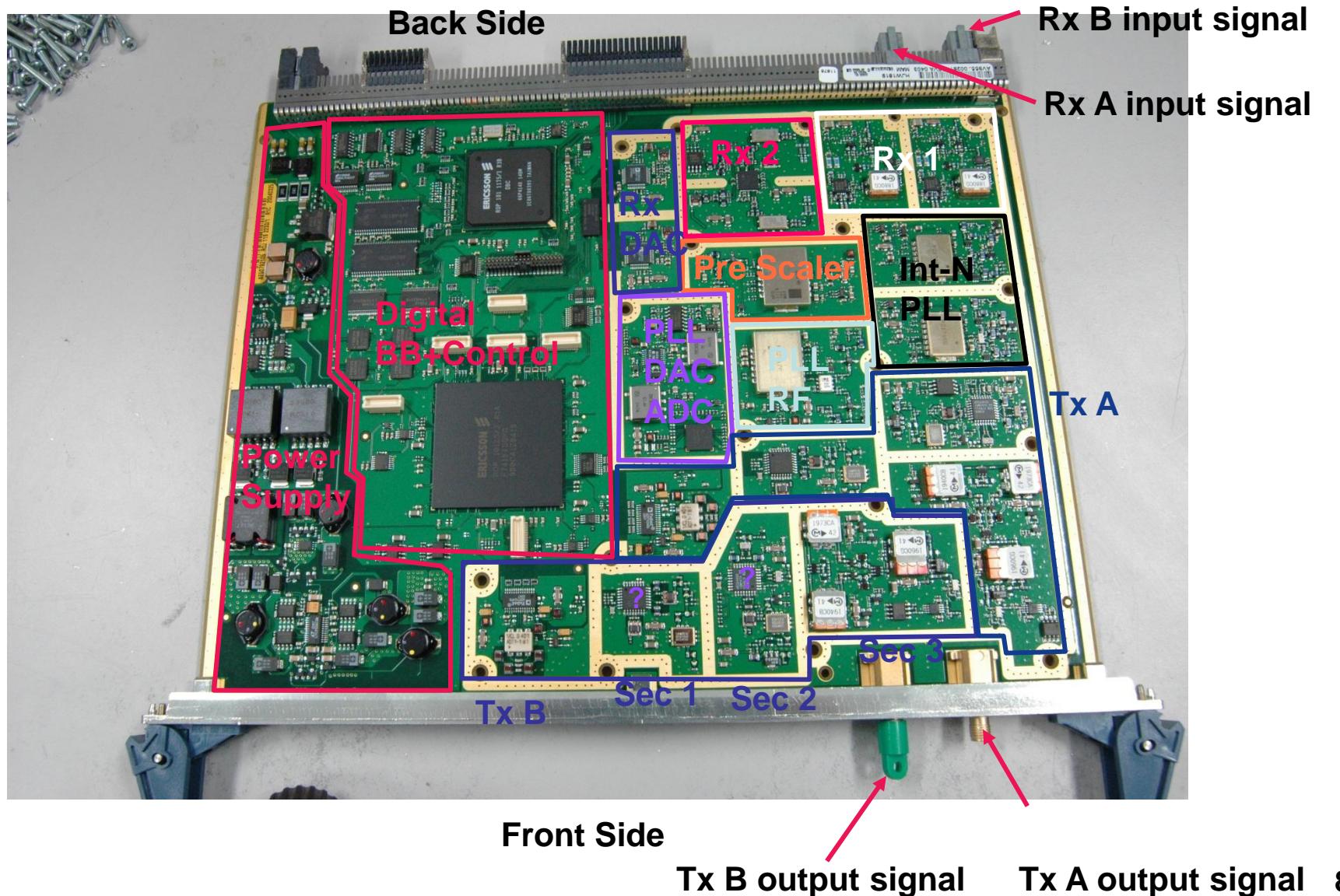


A single cabinet



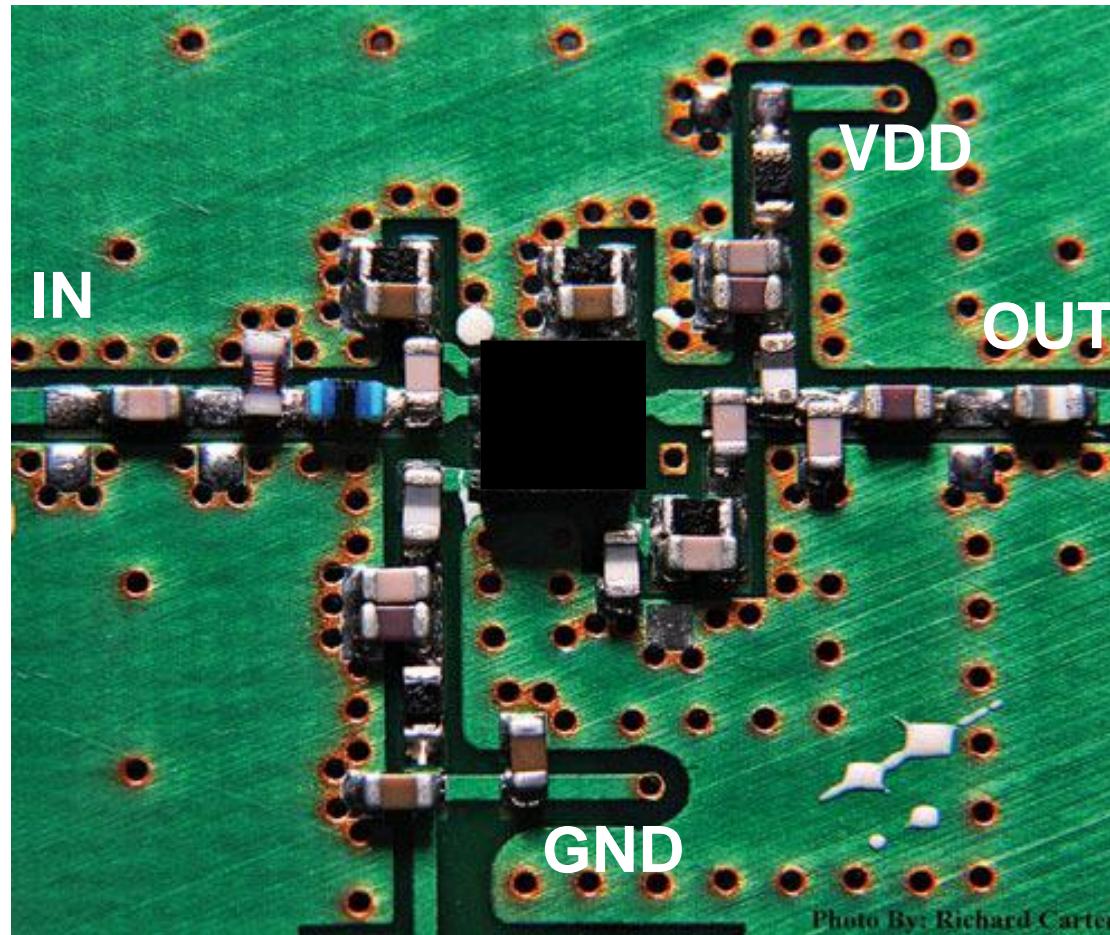
A single rack

# State-of-the-art BS Transceiver



# And a single RF IP block: LNA

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Rx LNA at 900MHz:  
22 SMD components around a single pHEMT GaAs device!

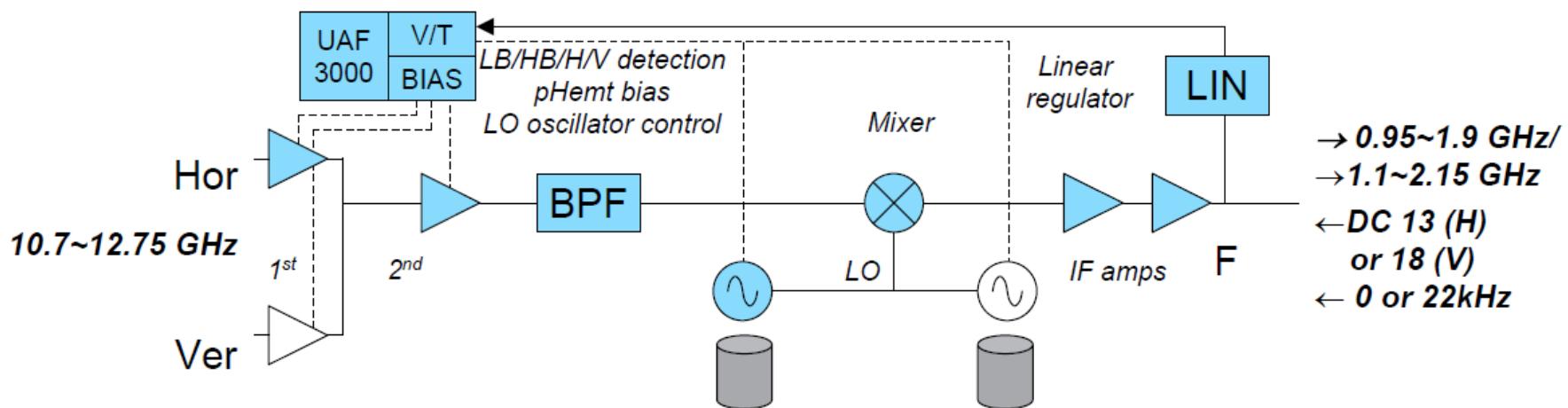
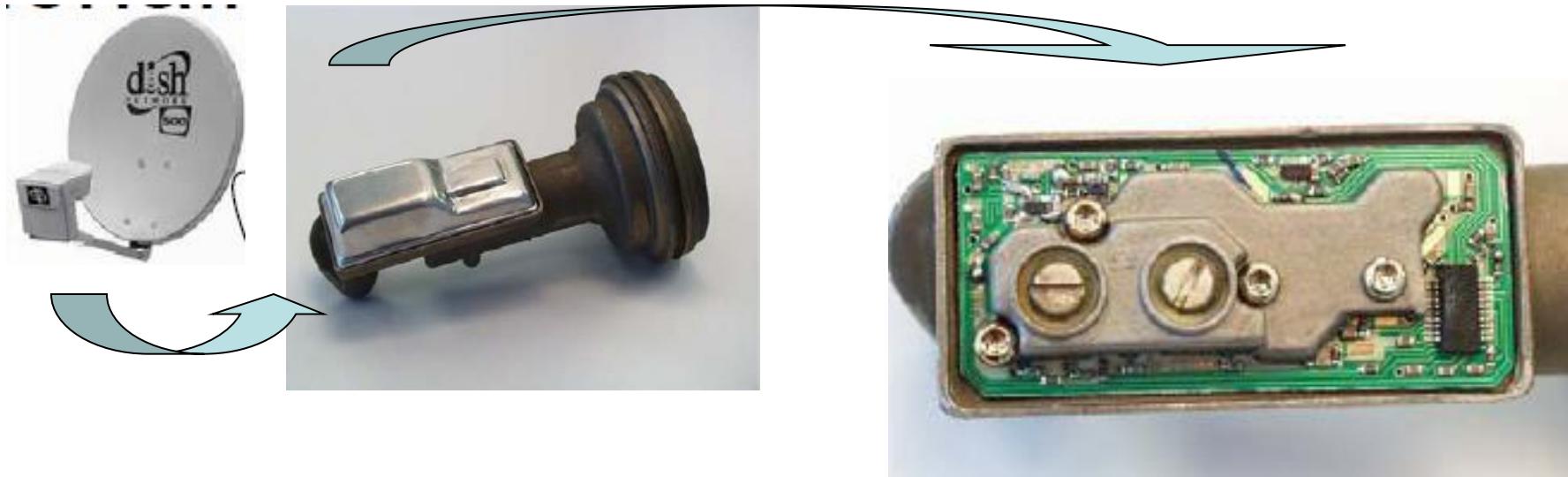
# Cellular base station

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- **Observation 1: Discrete components**
  - Expensive
  - Compound semiconductor technology
  - ...
- **Observation 2: Duplication**
  - Expensive

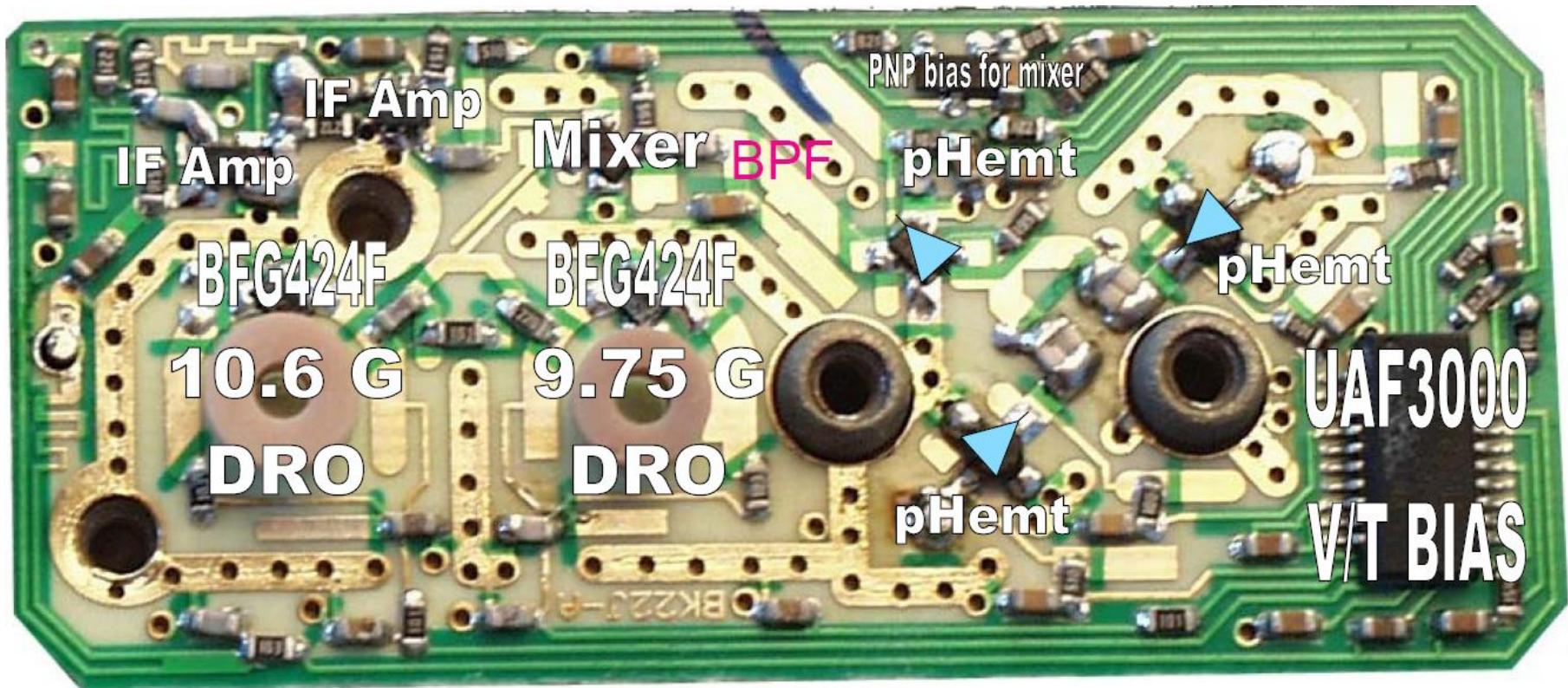
Why if we can reach impressive  
integration levels in the handheld?

# Satellite infrastructure: Ku band



Low Noise Block (LNB) down converter

# State-of-the-art Ku LNB



discrete implementation using mainly Compound Semiconductor Technologies (pHMET GaAs)

# Satellite infrastructure

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- **Observation: Discrete components**
  - Expensive
  - Compound semiconductor technology
  - Manual trimming
  - ...

Why is this down converter not integrated?

# Wireless infrastructure

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**it's performance!!**

**Cost is not an issue  
Size is not really an issue  
Power dissipation is not an issue**

# A few examples: base station LNA

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900MHz	Spec HH	Spec BS	avago	Asia-Pacific Microwave Conf. 2000	JSSC 2007
NF [dB]	< 2	< 0.6	0.53	1.35	0.2 non- $50\Omega$
1-dB OCP [dBm]	> -5	> +15	+18	+8	+1
OIP3 [dBm]	> 8	> +30	+33	+22	+12
Gain [dB]	> 15	> 15	18	13	17
S11 [dB]	< -10	< -20	< -20	< -10	< -11
Pdiss [mW]	low		216	38	43
technology			0.5μm GaAs	0.25μm BiCMOS	90nm CMOS

# A few examples: satellite LNA

Ku band

Ka band

10-12GHz	Spec	NEC	Ellinger 2004	Aspemyr 2006
NF [dB]	< 0.6	0.35	3	2.1
Gain [dB]	> 10	> 13	14	12.3
Pdiss [mW]		20	2.3	11.2
technology		0.25μm GaAs	0.25μm BiCMOS	90nm CMOS
18-22GHz	Spec	NEC	Guo 2005	
NF [dB]	< 1	0.7	5.5	
Gain [dB]	> 10	> 13	10	
Pdiss [mW]		20	24	
technology		0.25μm GaAs	0.13μm CMOS	

# A few examples: satellite VCO

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15GHz	Hittite	Asia-Pacific Microwave Conf. 2009	BCTM 2007
Frequency	15GHz	13.5GHz	15GHz
PN @ 100kHz [dBm/Hz]	-105	-84.4	-82
Po [dBm]	+7	-4	+2
Pdiss [mW]		4.5	18
technology	GaAs - InGaP	0.18μm CMOS	0.18μm SiGe:C

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# The need for integration is coming

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- In cellular base stations
  - Increased density of base stations to cope with data traffic (Shanghai: every 500m!)
  - More standards / bands / phased arrays
  - Power reduction: the ‘green’ label
  - Tower mounted units instead of ‘base’ station
  - Robustness, yield, ...

# The need for integration is coming

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- In satellite communication
  - Next Ka-satellite market India and South Americas: but then price of LNB must reduce
  - LNB/BUC is powered by Set Top Box (STB)
    - Ultimate goal is the universal QUAD LNB/BUC supplying 4 STBs minimum, but powered by still 1 STB worse case (backward compatible)
      - Ku-band & Ka-band & both polarizations

# **The need for integration is coming**

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- **Wireless infrastructure will benefit from integration**
  - More functionality
  - Improved yield and thus lower cost
  - Improved reliability (ESD, automatic calibration, BIST, adaptive matching, ...)
  - less PCB issues at costumer site
  - Paves the path towards spatial selectivity to mitigate interference, support tracking, boost TX power, etc.
- **But this all requires Si-based RF solutions**
  - GaAs-like technology lack dense integration

# Wireless infrastructure

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**so we need Si solutions**

**Two-step approach:**

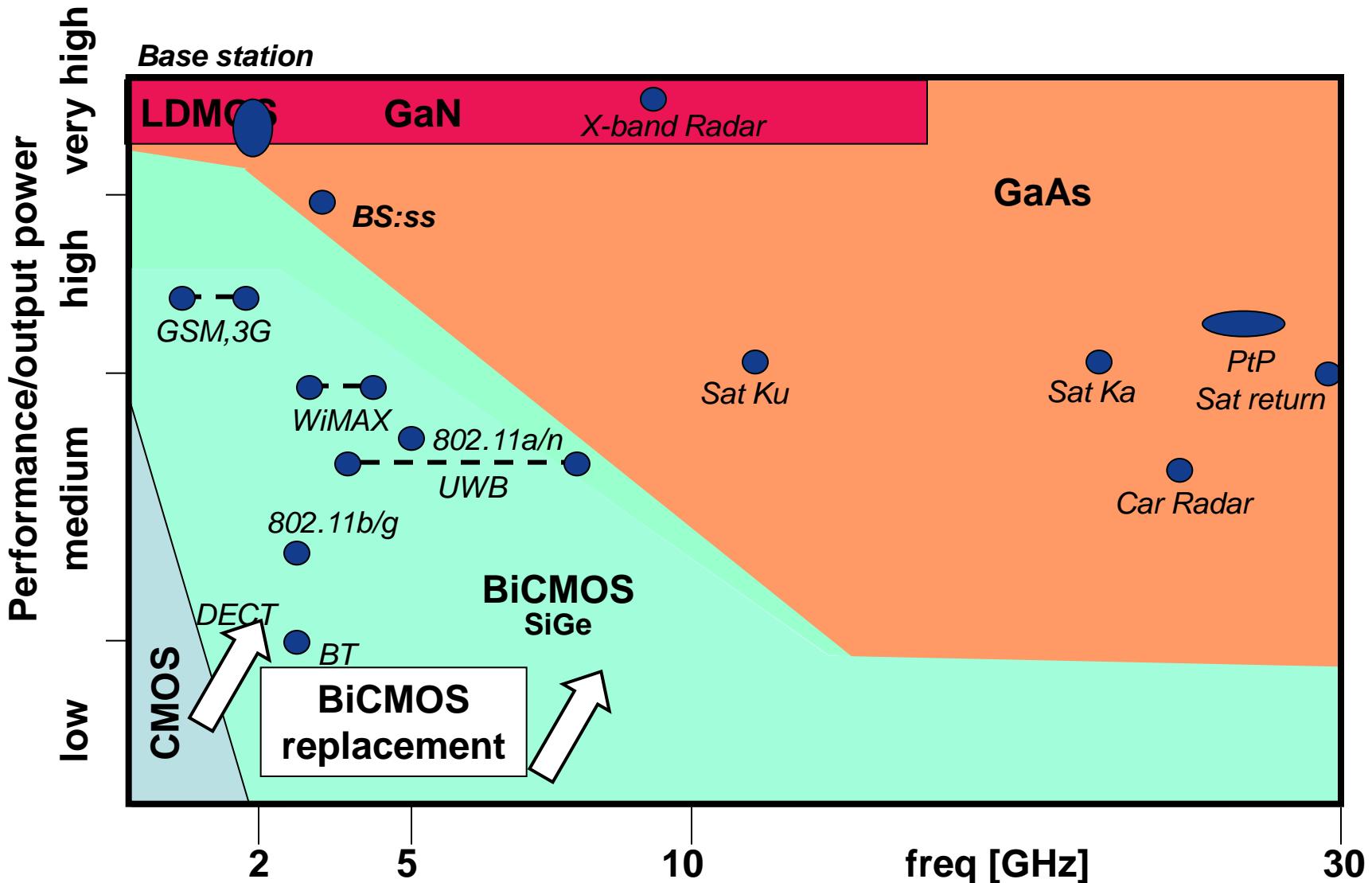
**Improve Si-technology  
Different architecture/circuit design**

# Content

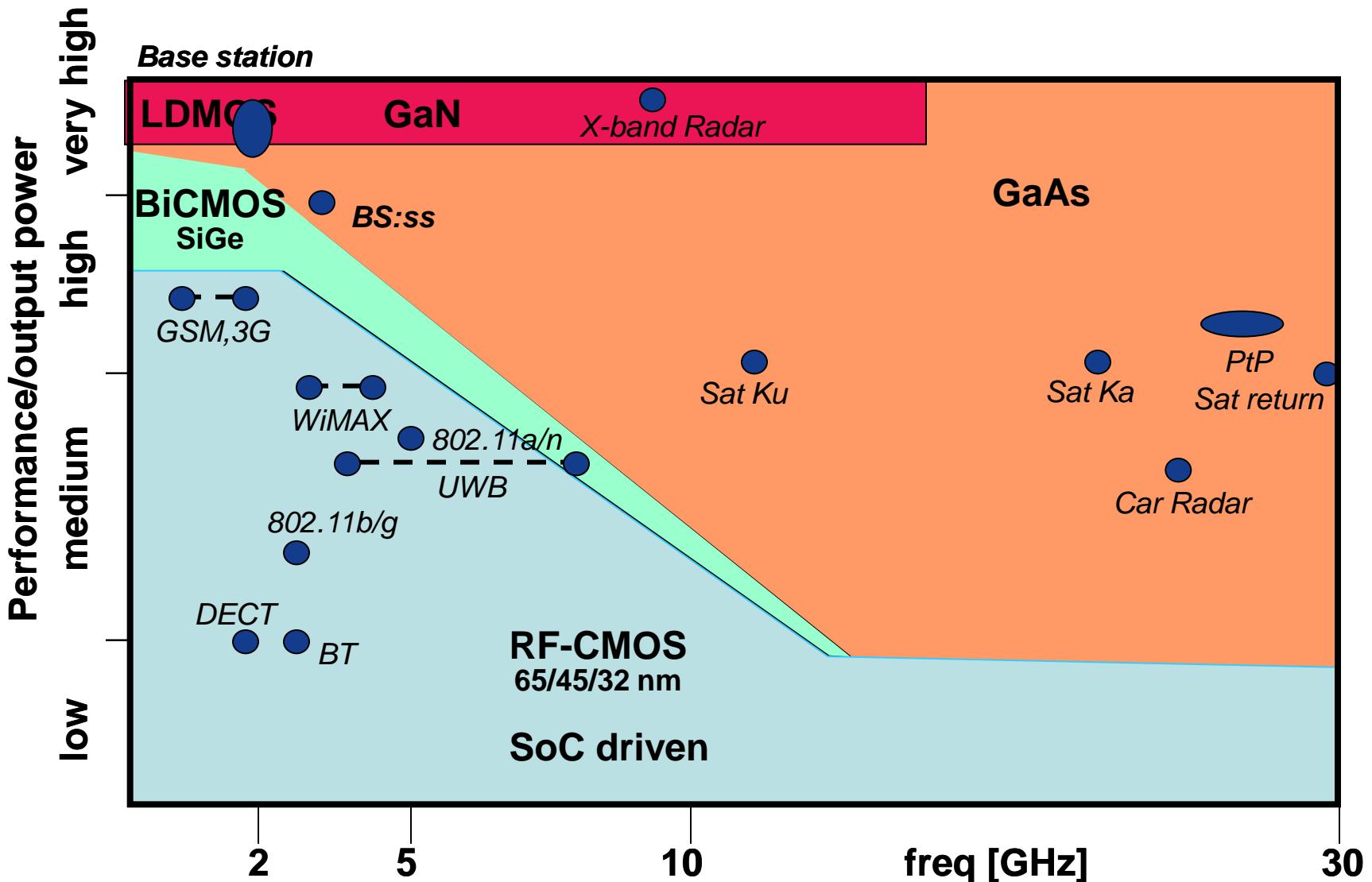
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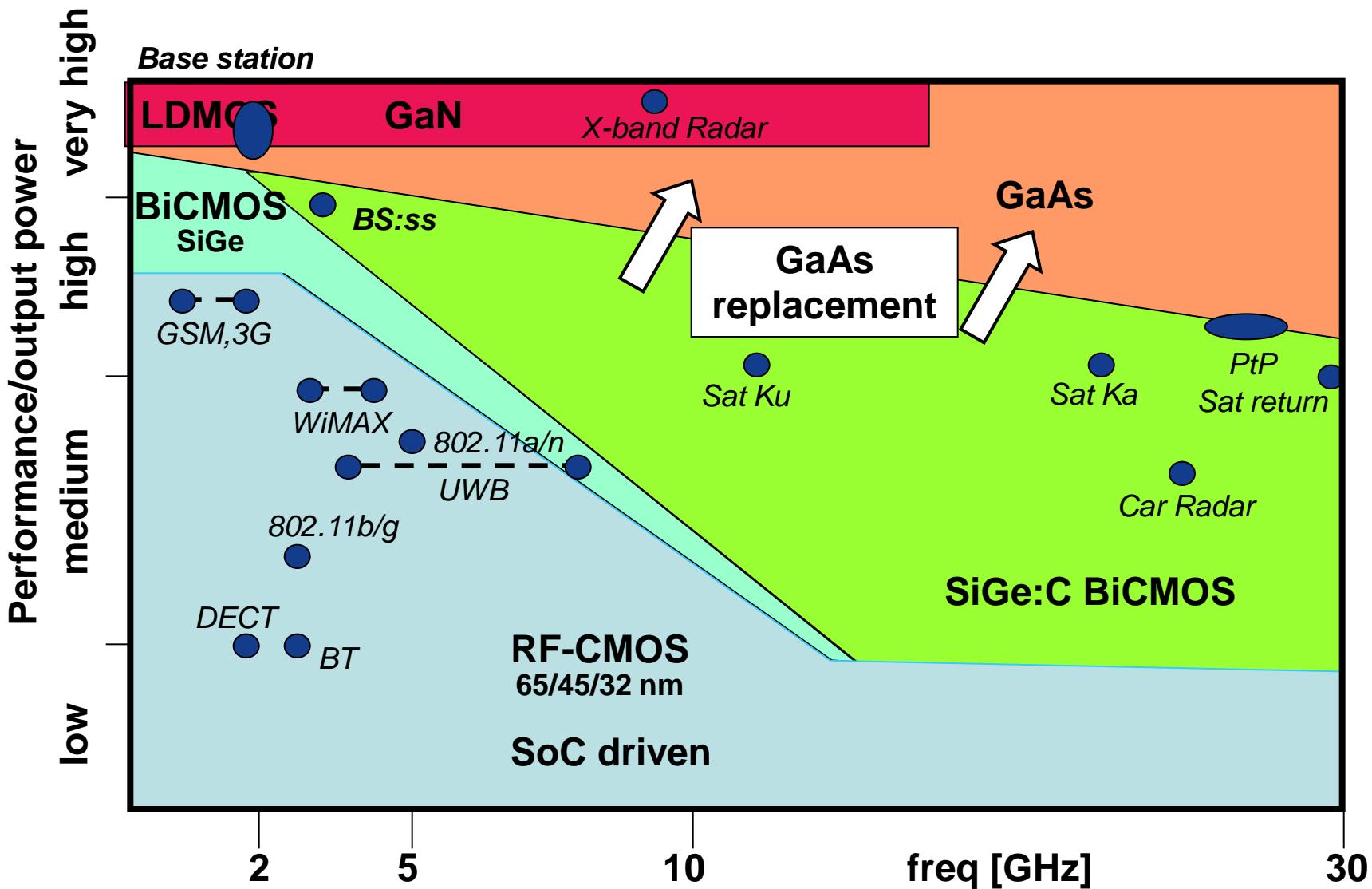
# Improve Si-technology



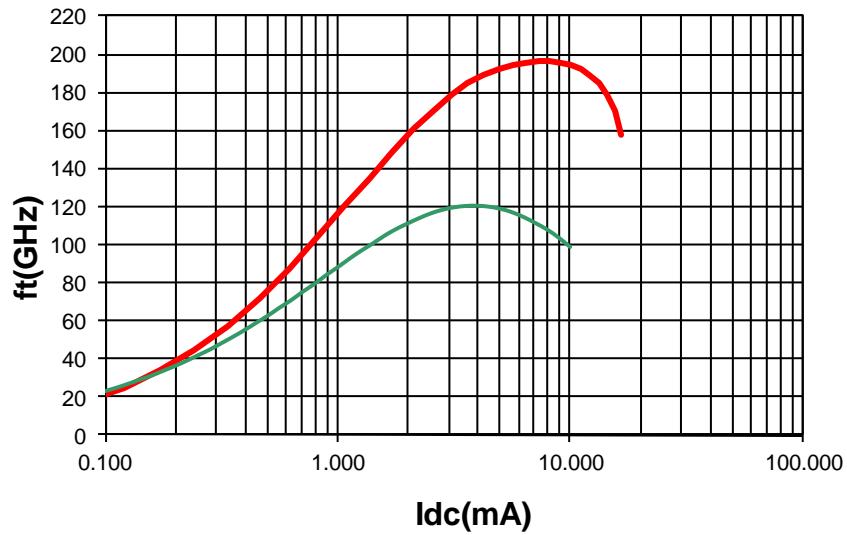
# Improve Si-technology



# Improve Si-technology

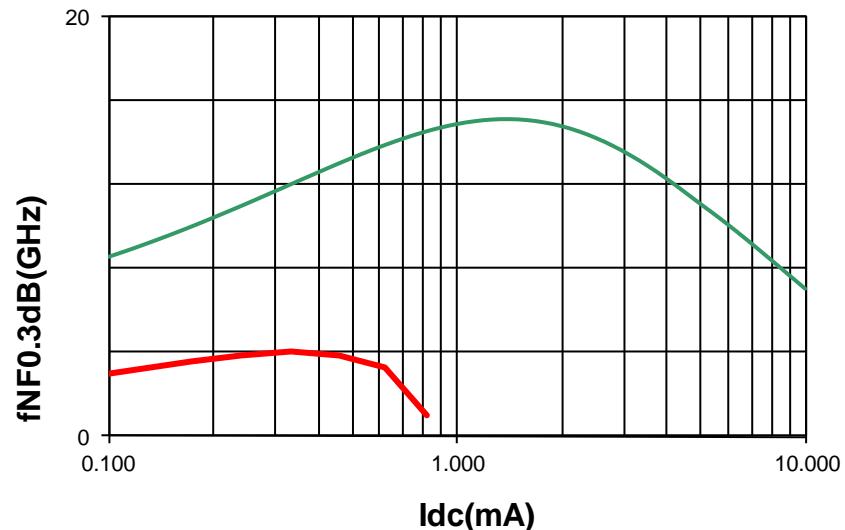


# Technology: SiGe:C front-end



**SiGe:C NXP technology**  
**TSMC 65nm CMOS**

65nm NMOS → capacitive nature  
SiGe:C → real part (base resistance)



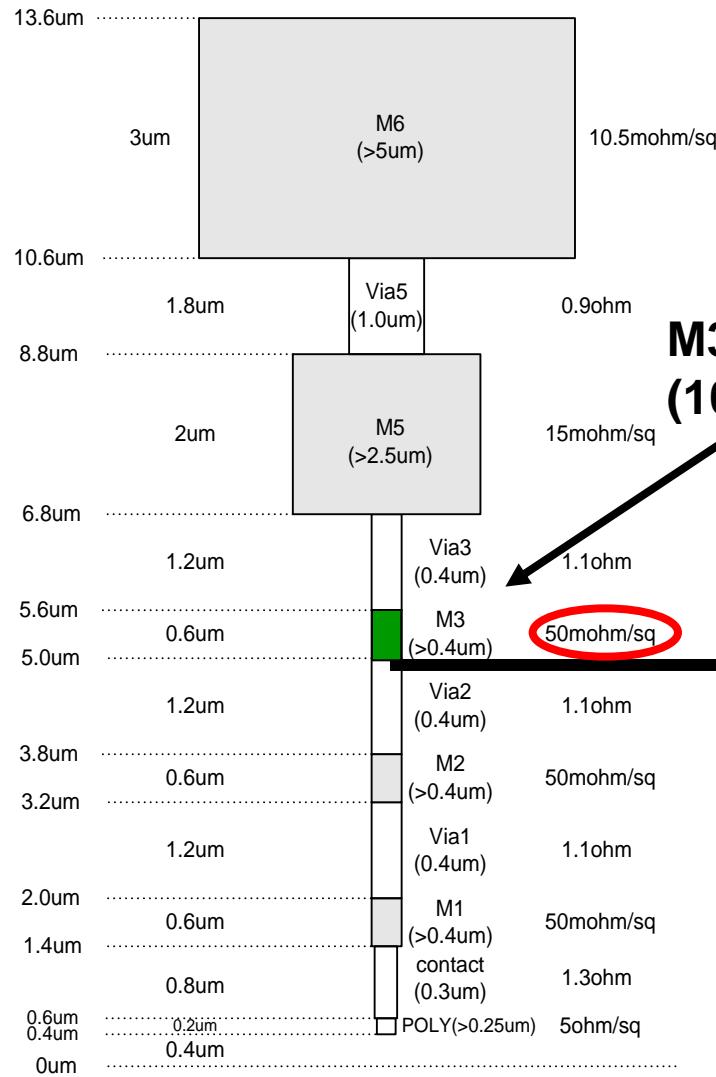
**Impedance match**

$$Z_s = Z_i^* \Rightarrow P_{s,av} = P_{i,del}$$

**Noise match**

$$Z_s = Z_{opt}$$

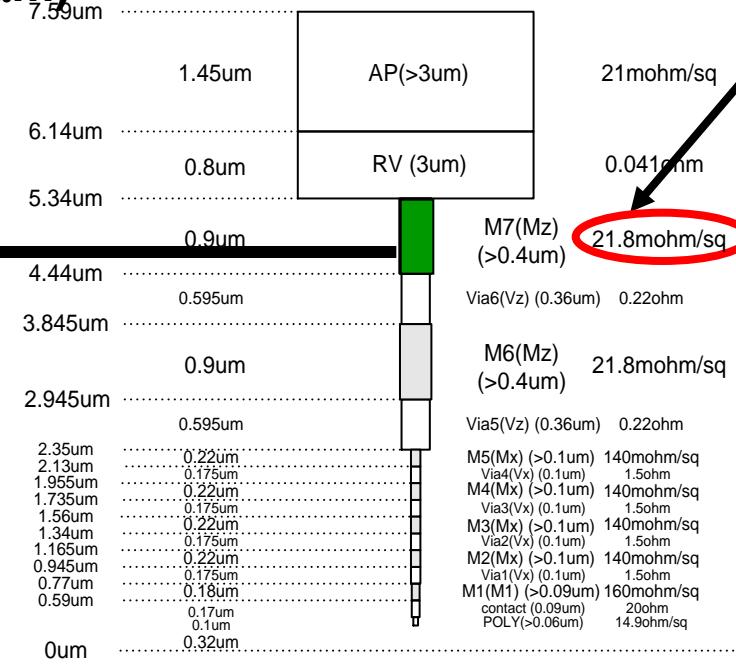
# Technology SiGe:C: back-end



**superior RF performances**

**M3-SUB: 0.65fF  
(10μm/0.4μm)**

**M7-SUB: 7.4fF  
(10μm/0.4μm)**



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# **High performance RF design**

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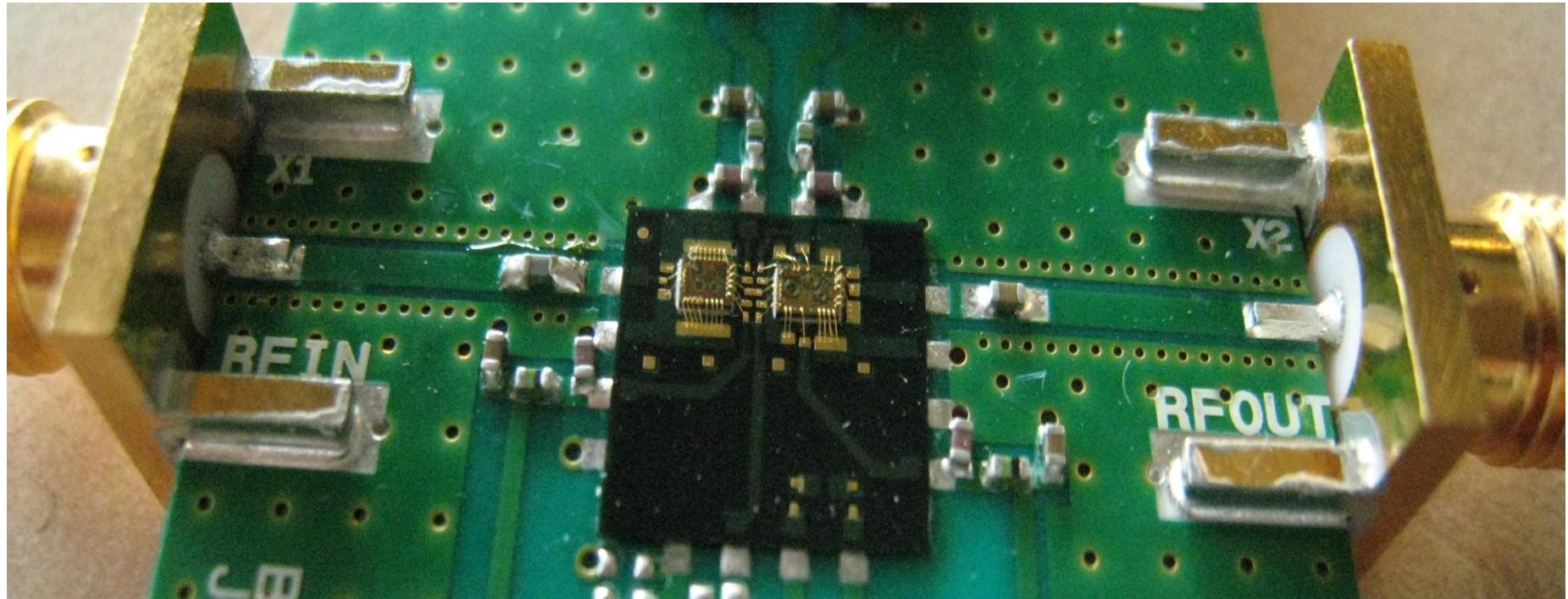
**Use the density of Si-based technology**

- Use advanced architectures
- Use digital!
  - Adaptivity
  - Calibration
  - Signal processing power
  - ...

**and a mix of disciplinary fundamental  
research (IC design, EM, SP, ...)**

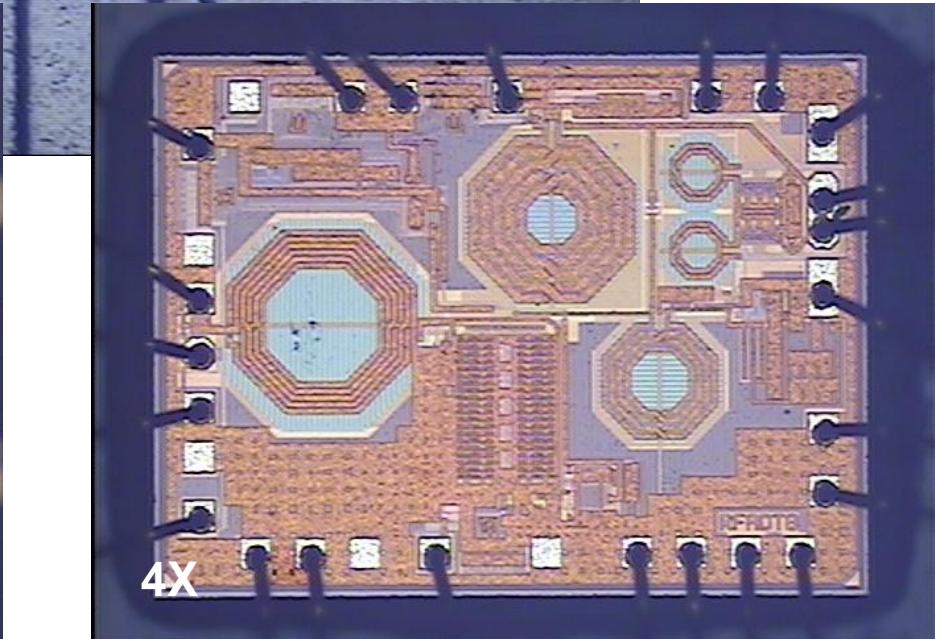
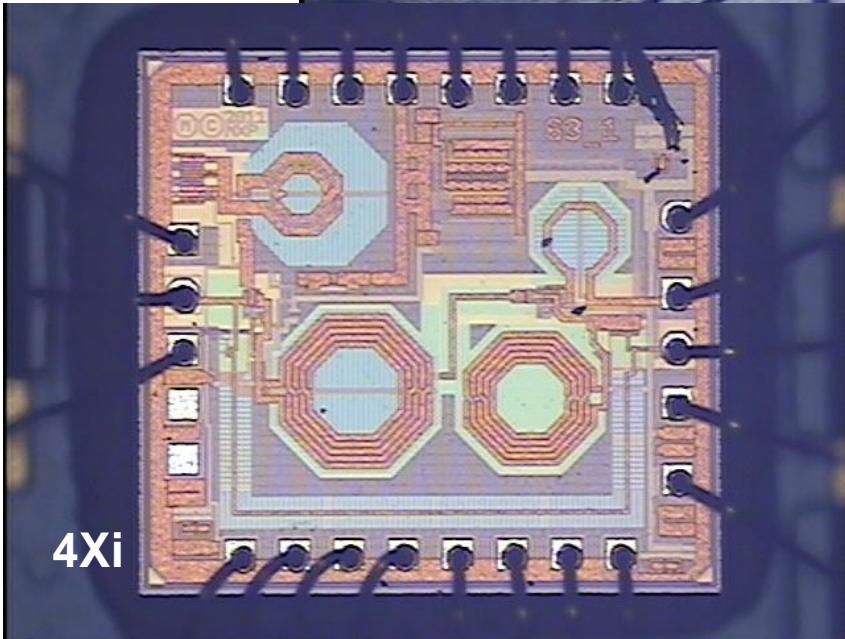
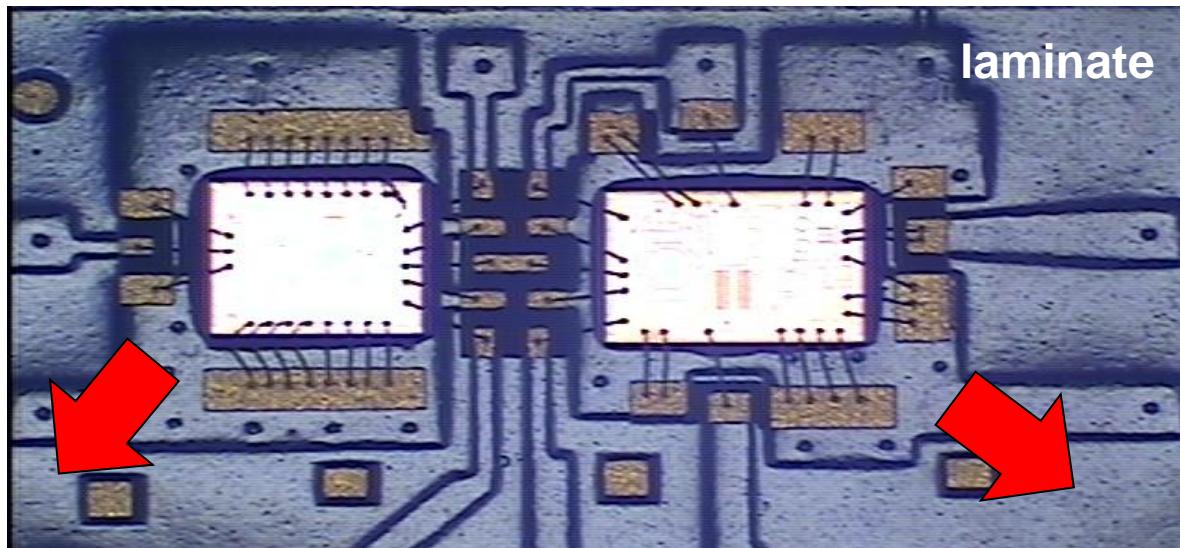
# An example: BST LNA

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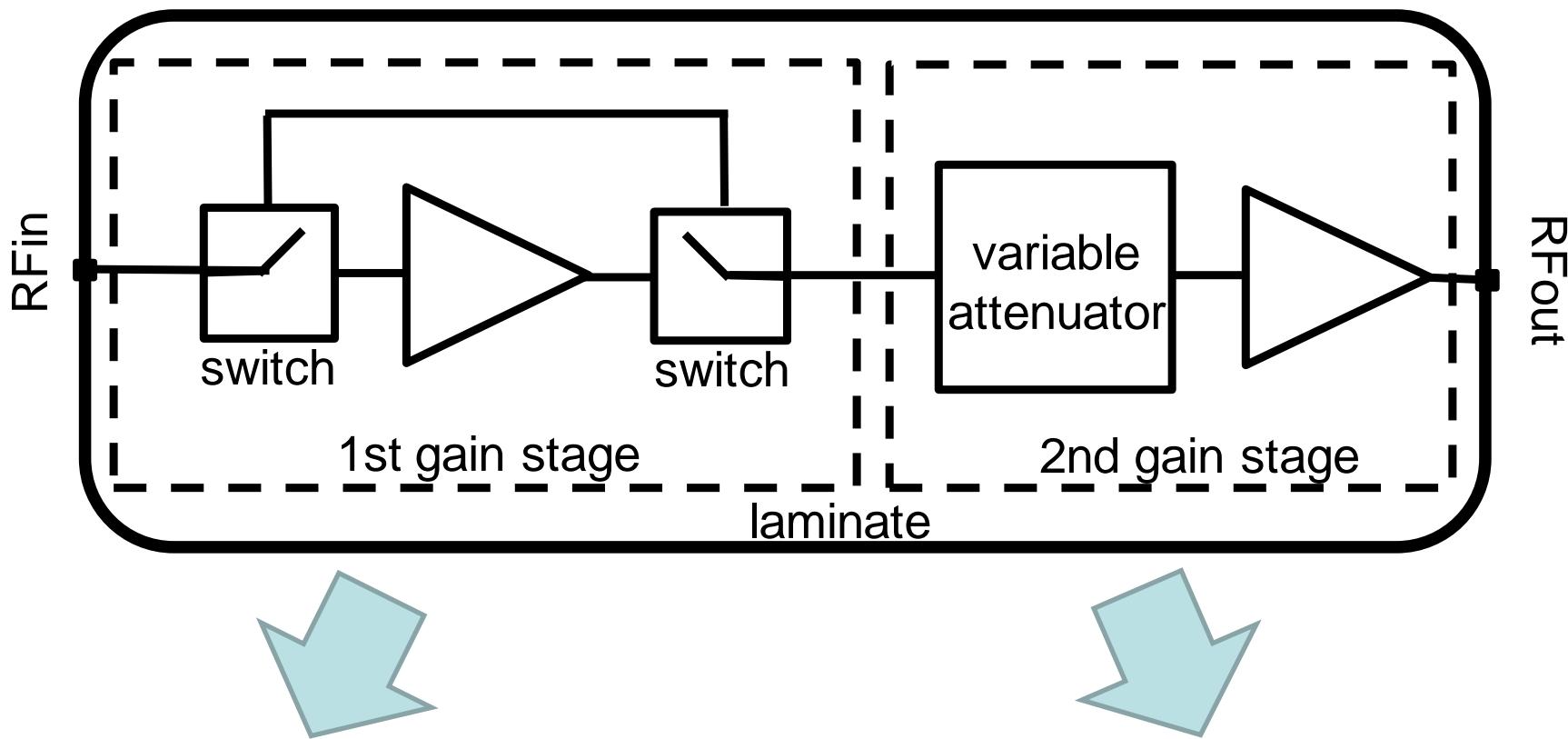


Package: HLQFN16R

# Die photo's



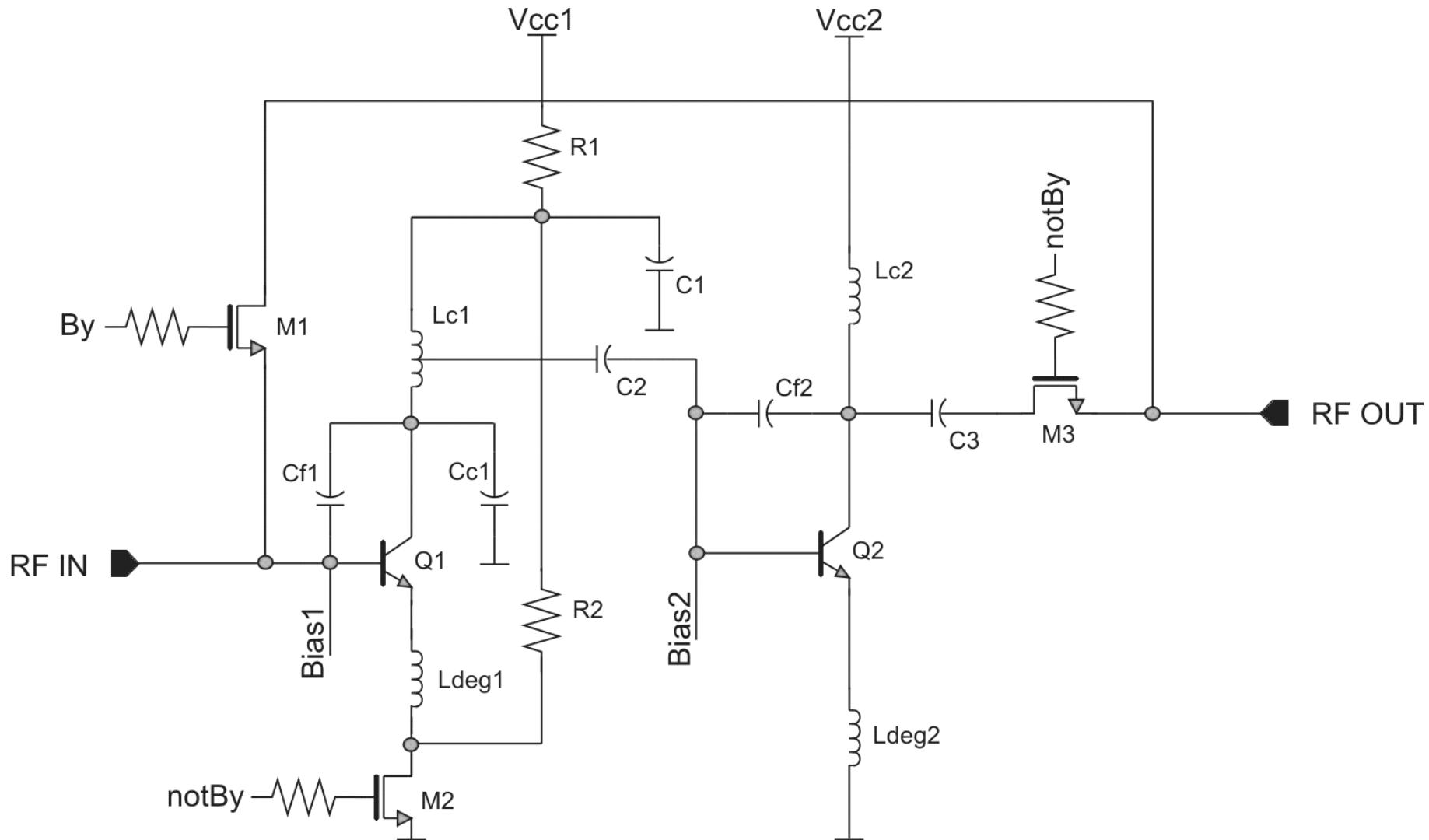
# Why a Module?



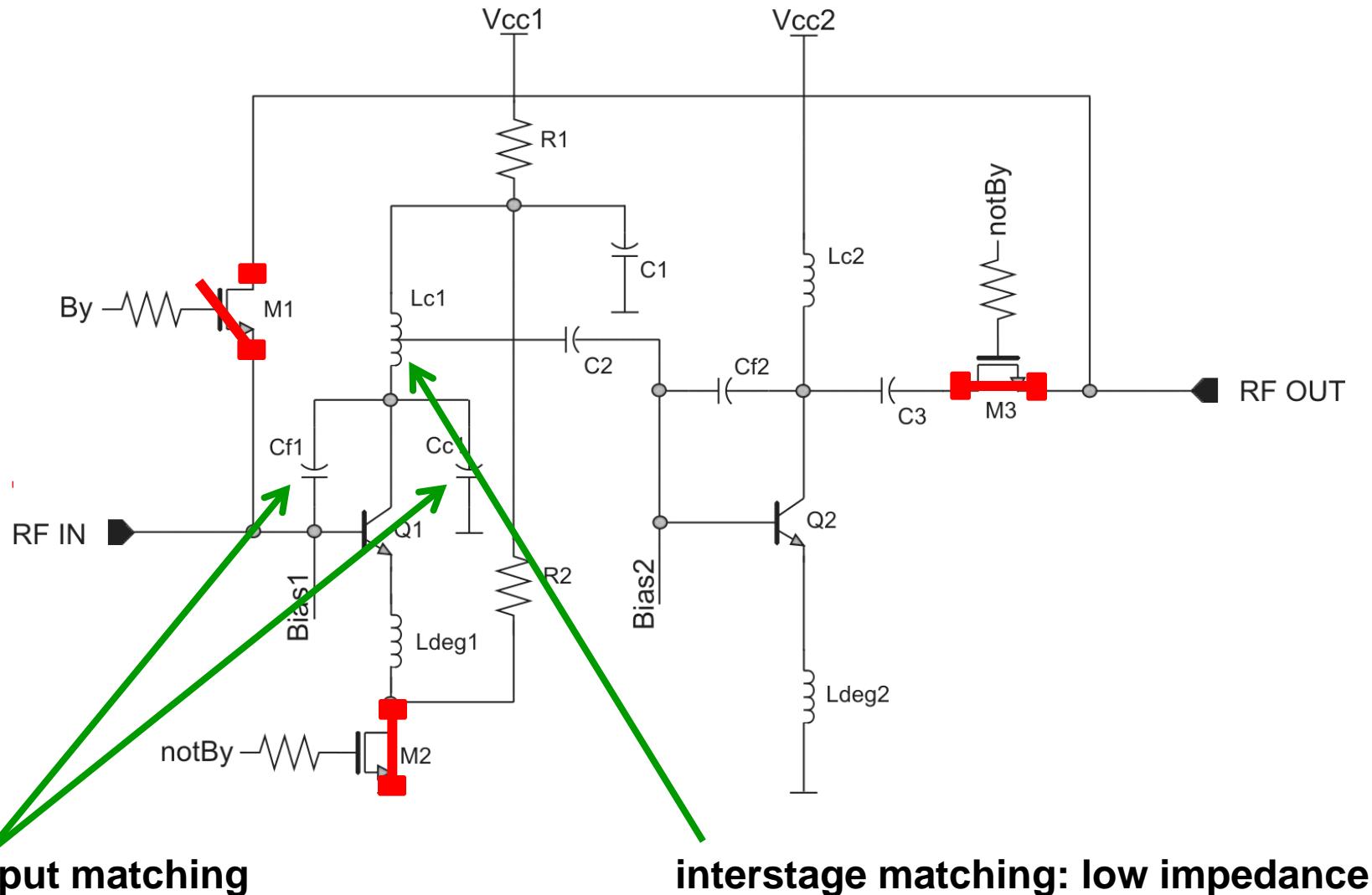
0.25 $\mu$ m SiGe:C BiCMOS  
ft = 216GHz, BVceo = 1.45V  
NFmin (@2GHz)  $\approx$  0.4dB

0.25 $\mu$ m SiGe:C BiCMOS  
ft = 130GHz, BVceo = 3V  
NFmin (@2GHz)  $\approx$  0.7dB

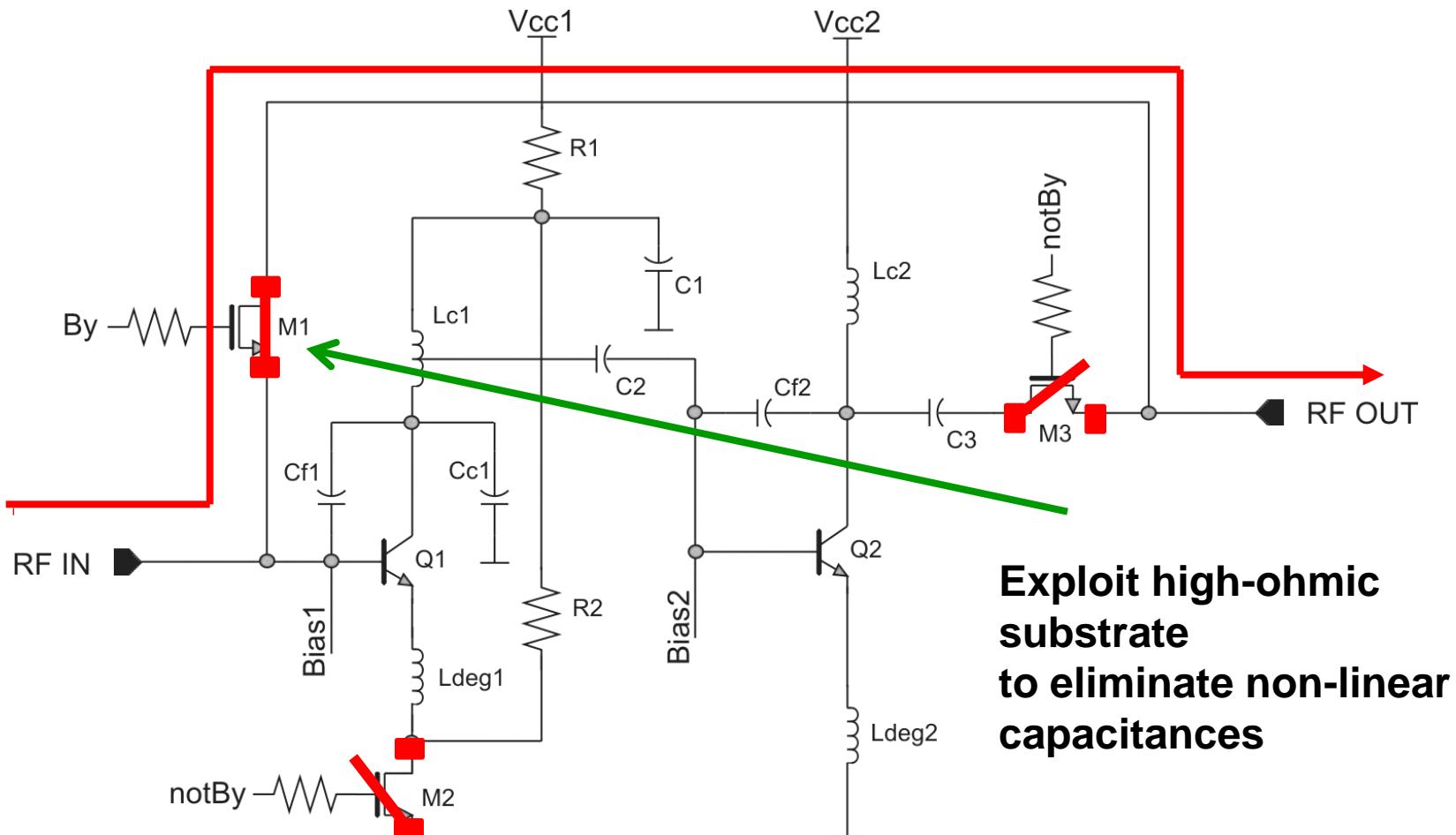
# Design: 1<sup>st</sup> gain stage



# 1<sup>st</sup> gain stage: gain mode

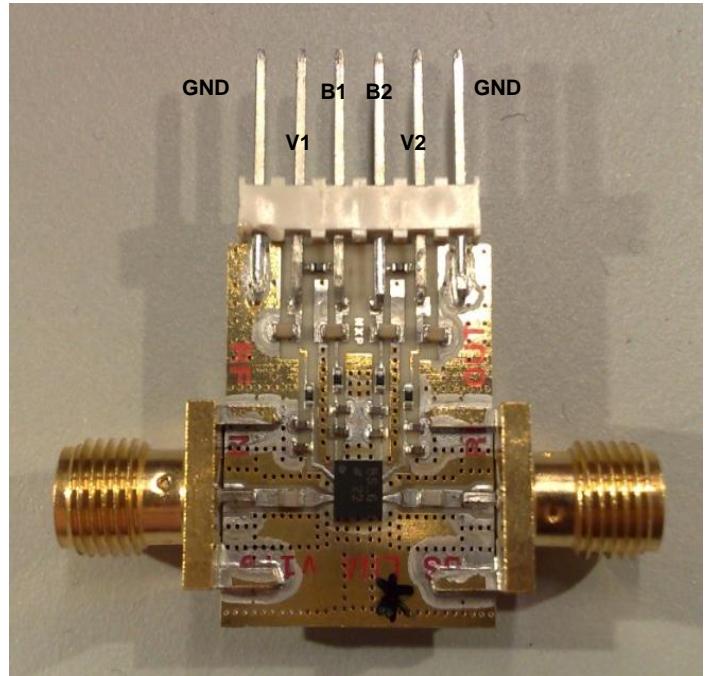
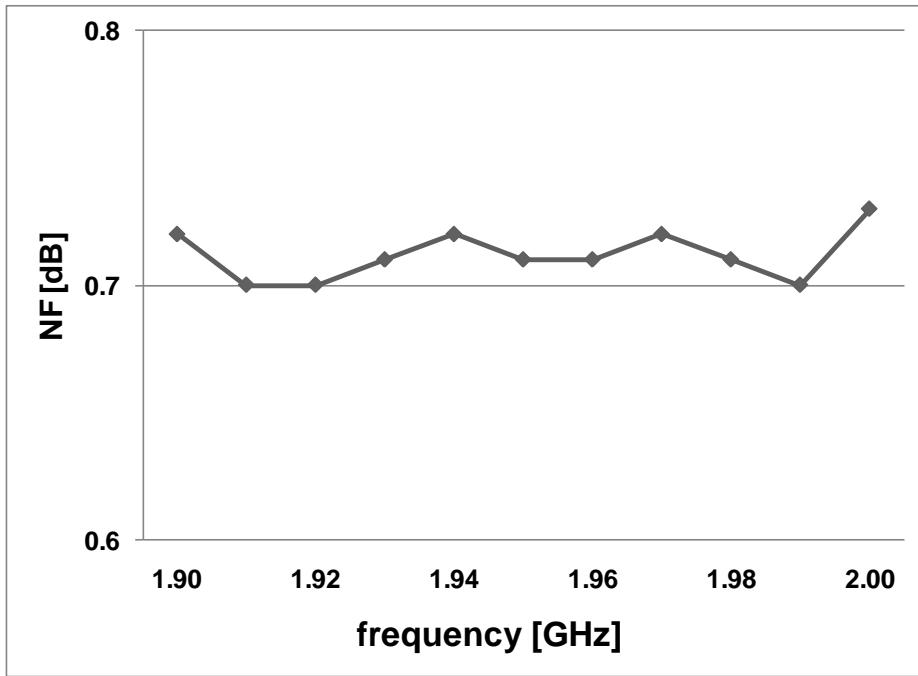


# 1<sup>st</sup> gain stage: bypass mode



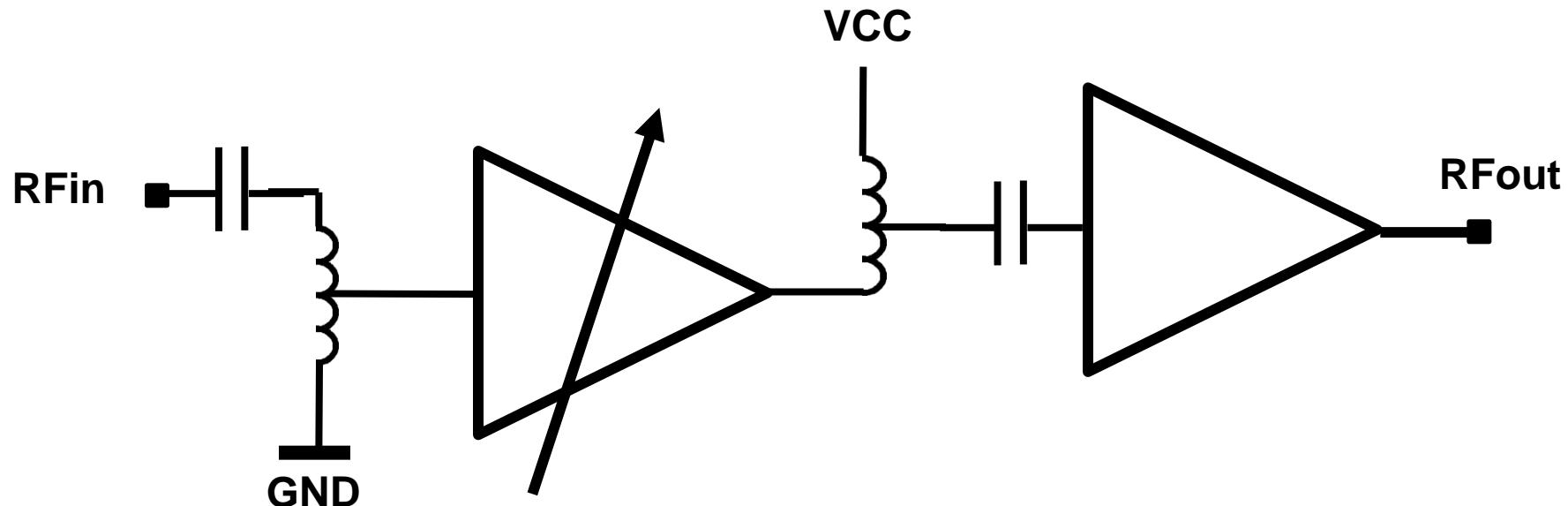
**Note: 2nd die will influence the input matching!!**

# Measurement 1<sup>st</sup> gain stage



Performance	Active mode	Bypass mode
Pdiss [mW]	350	0
Gain [dB]	19	-1.1
Input IP3 [dBm]	+14	+41
Output IP3 [dBm]	+33	+40
Output CP1dB [dBm]	+15	>+15

# Design: 2<sup>nd</sup> gain stage



input matching

variable attenuator

output stage

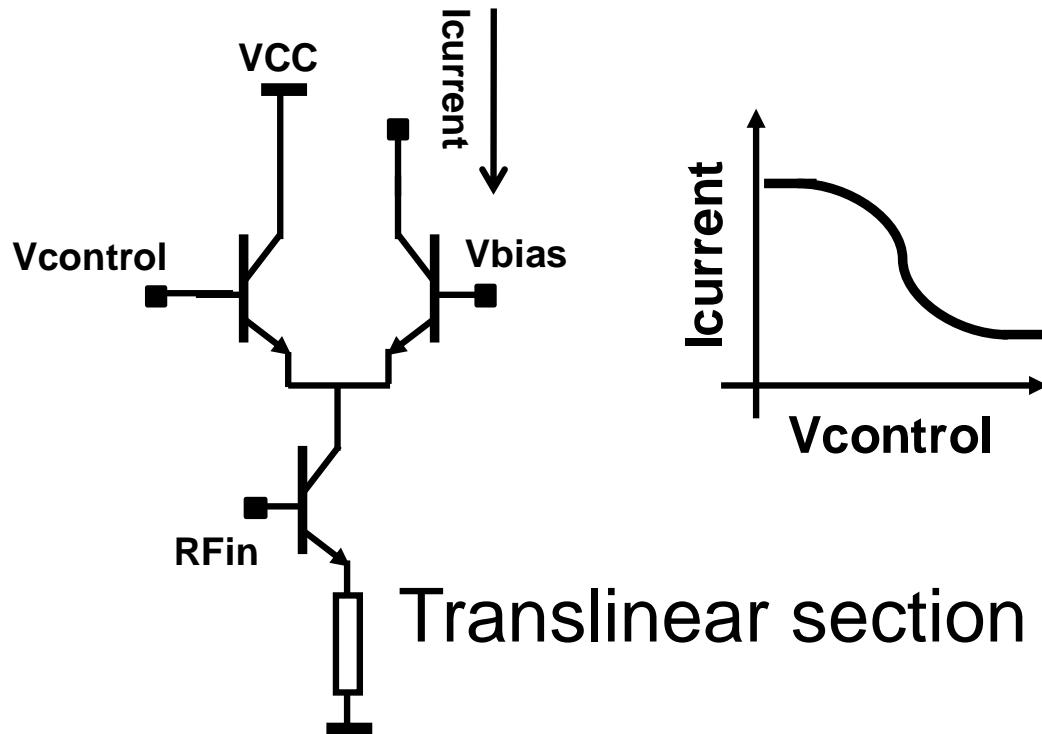


0 – 30 dB



High OIP3

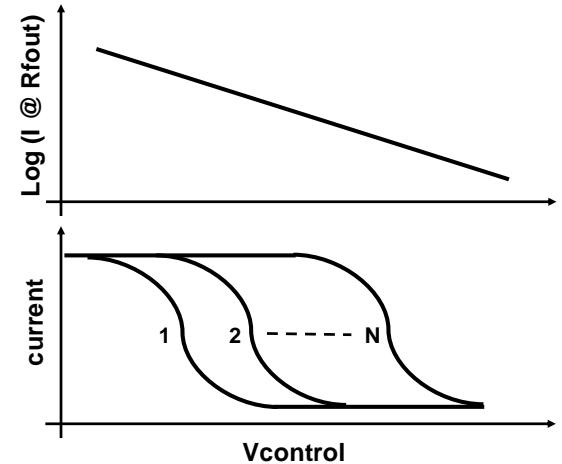
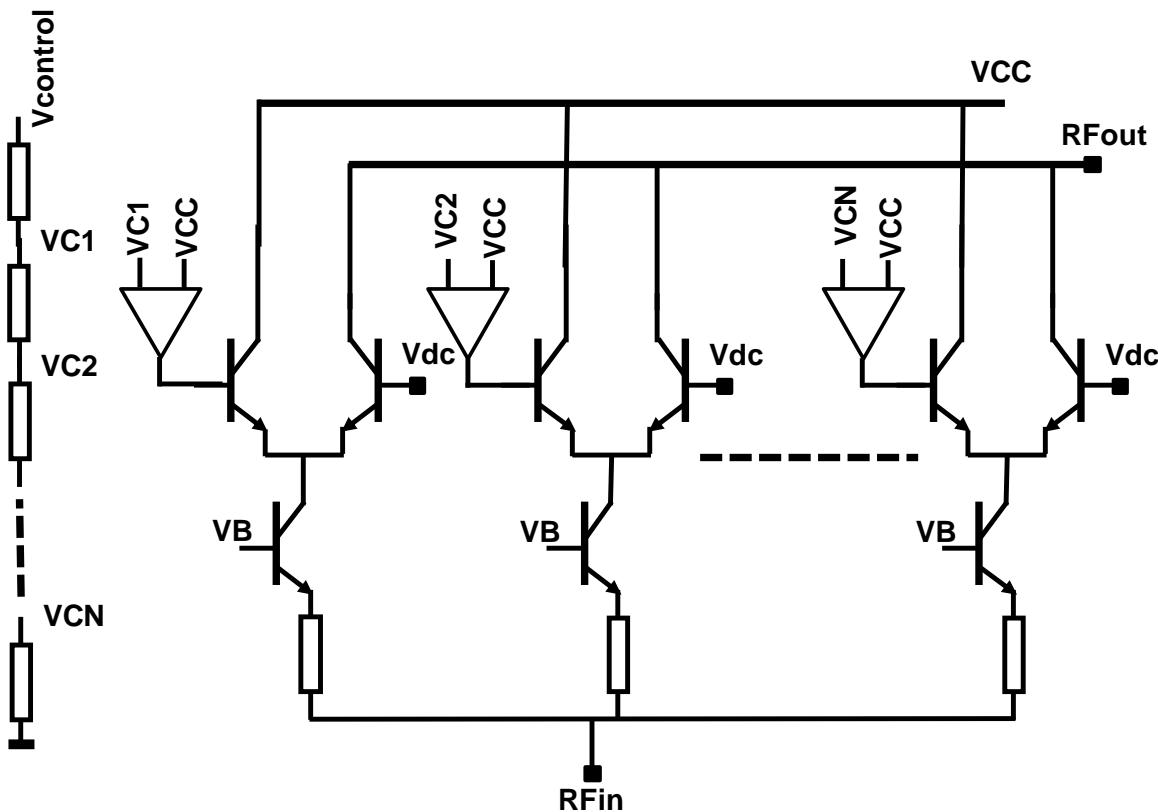
# Design: variable attenuator



Translinear section instead of MOS switches

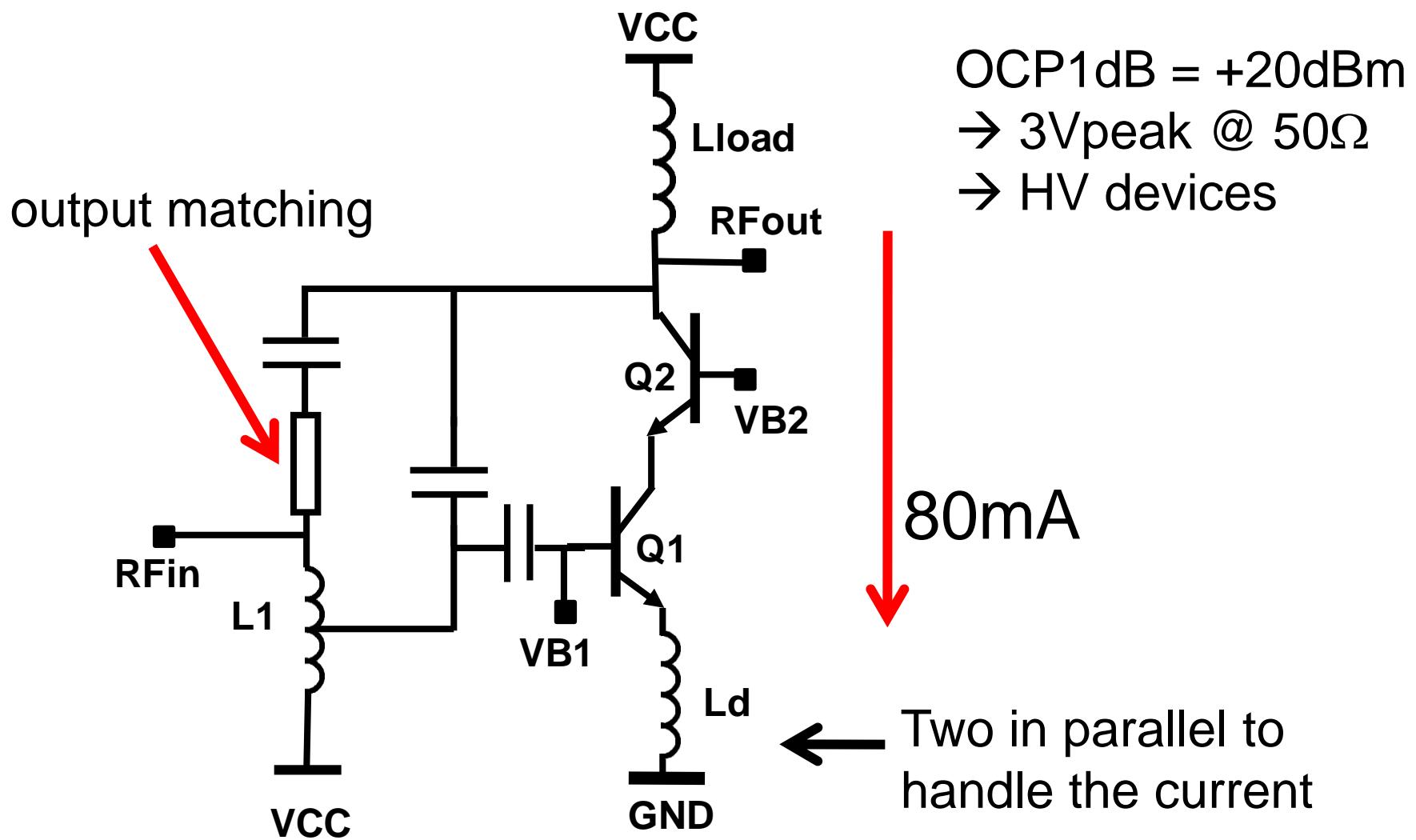
- better linearity
- very good reverse isolation
- slightly higher NF
- gain

# Design: variable attenuator



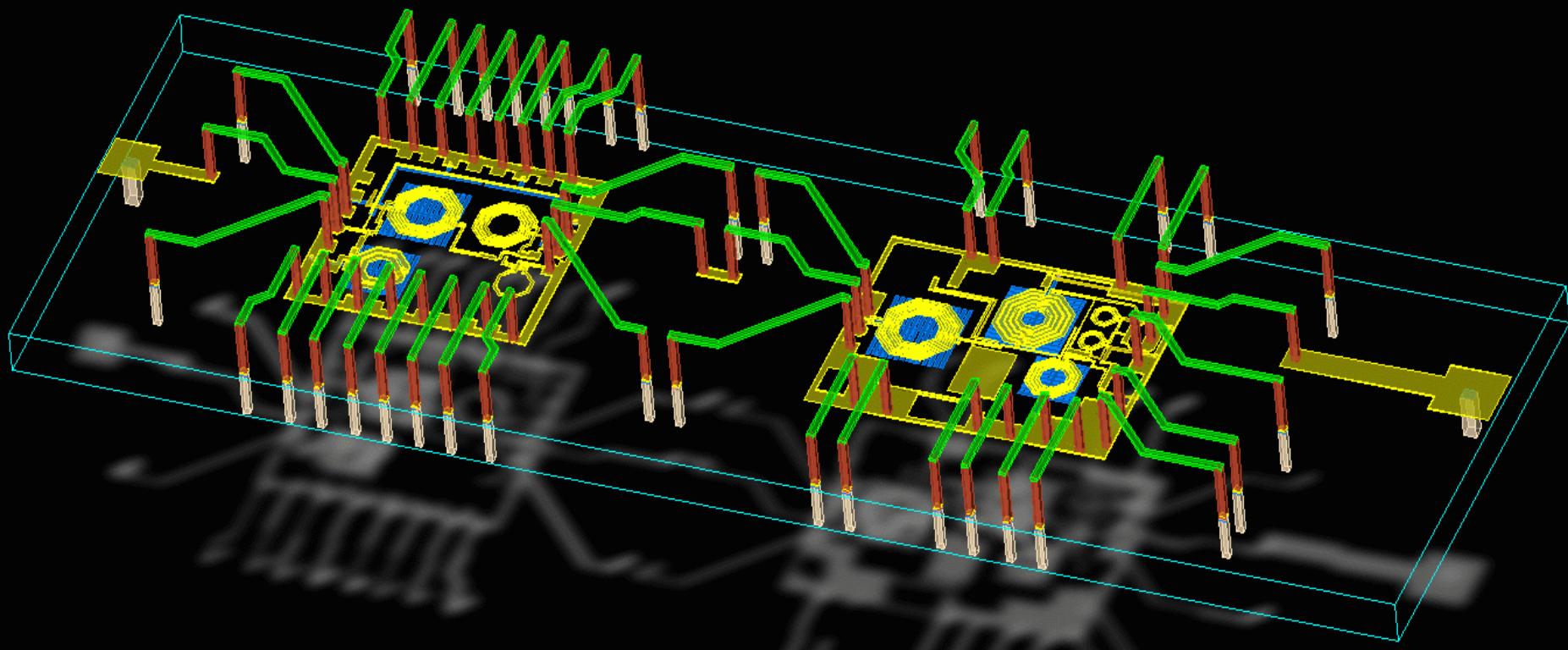
	JSSC 2008	JSSC 2011	This work
NF	3.5	1.5	6
IIP3	22	17	32

# Design: output stage



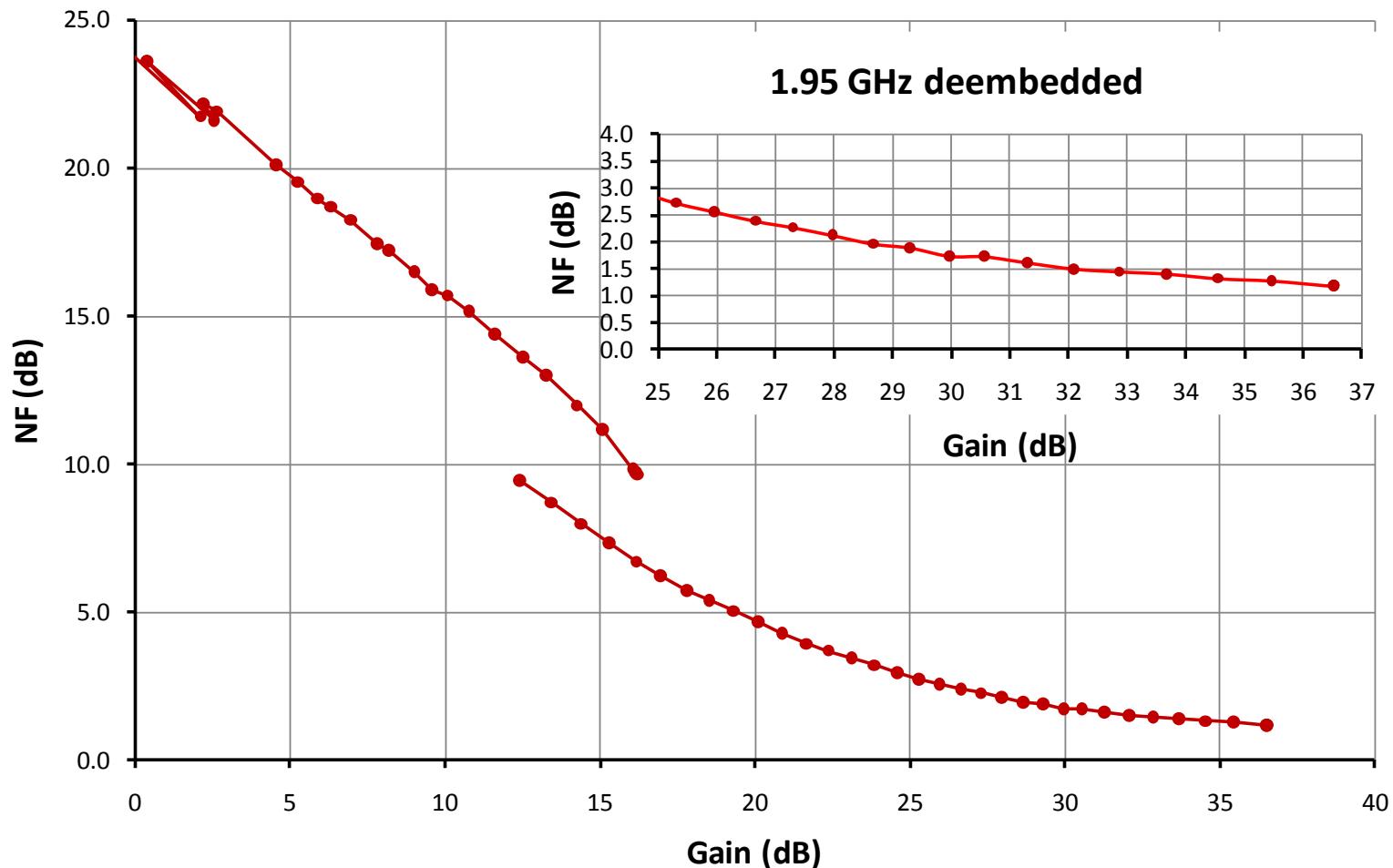
# Design: 2 dies

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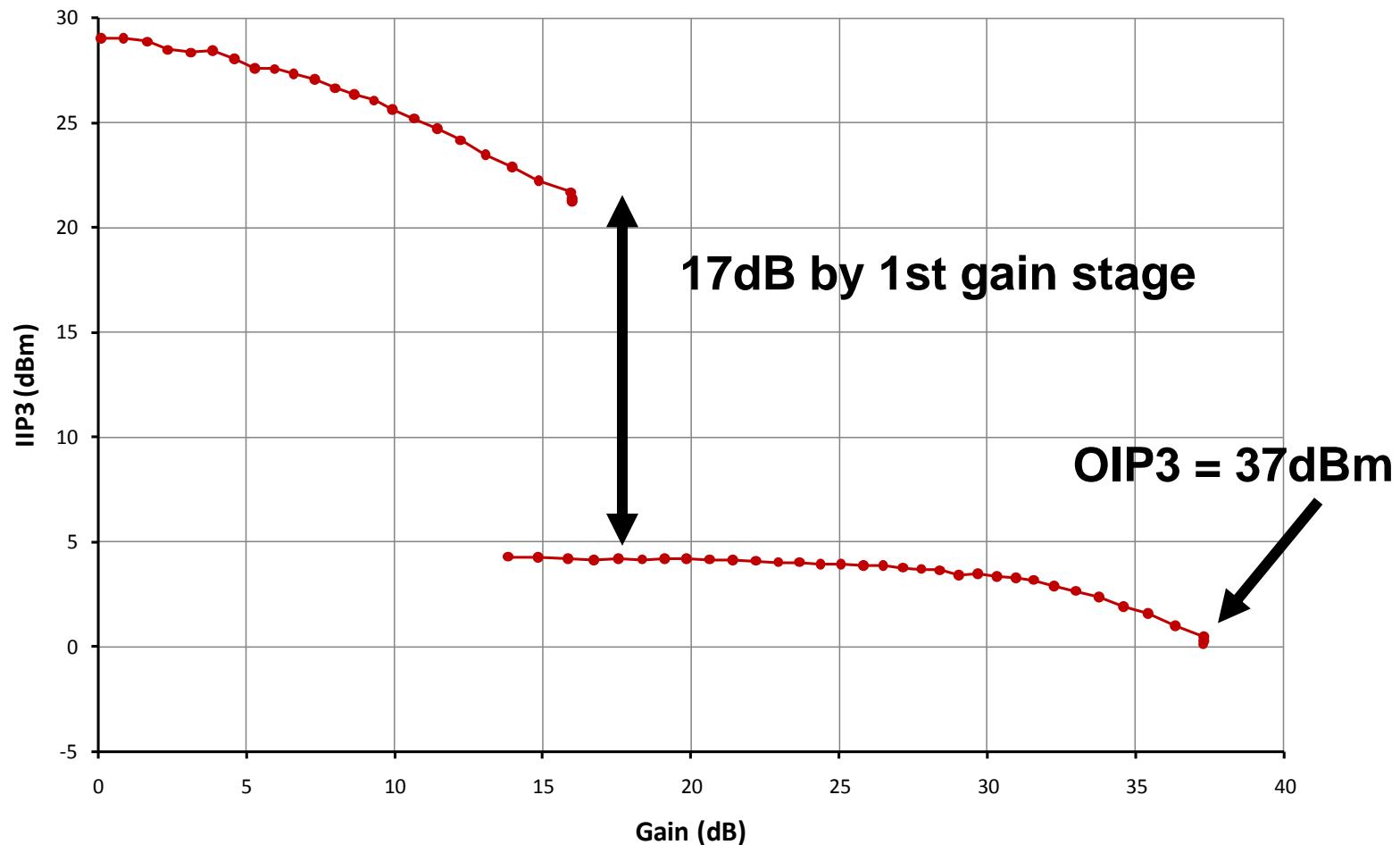
z  
y  
x

# Measurements: NF vs. Gain



Measurements performed at 65°C

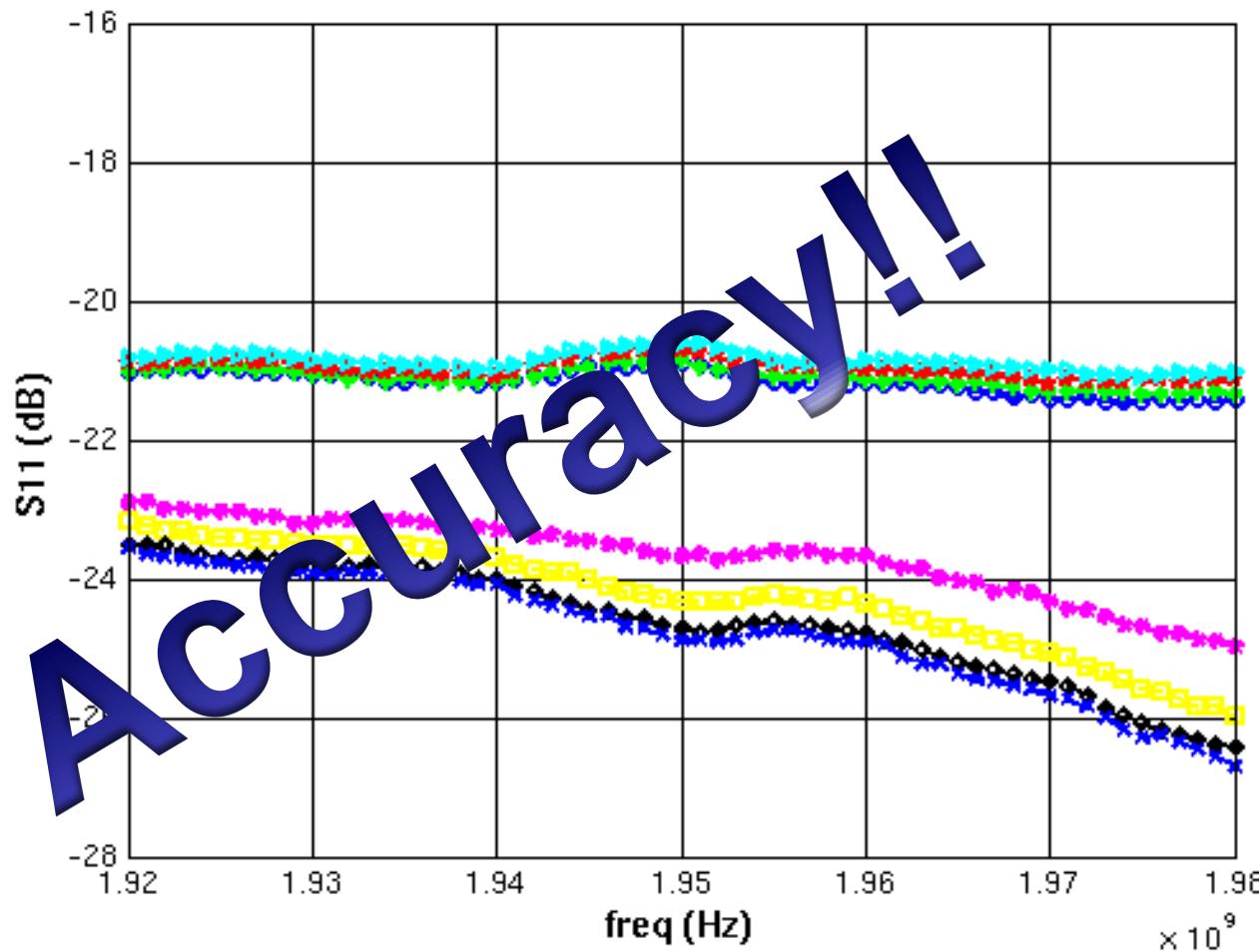
# Measurements: IIP3 vs. Gain



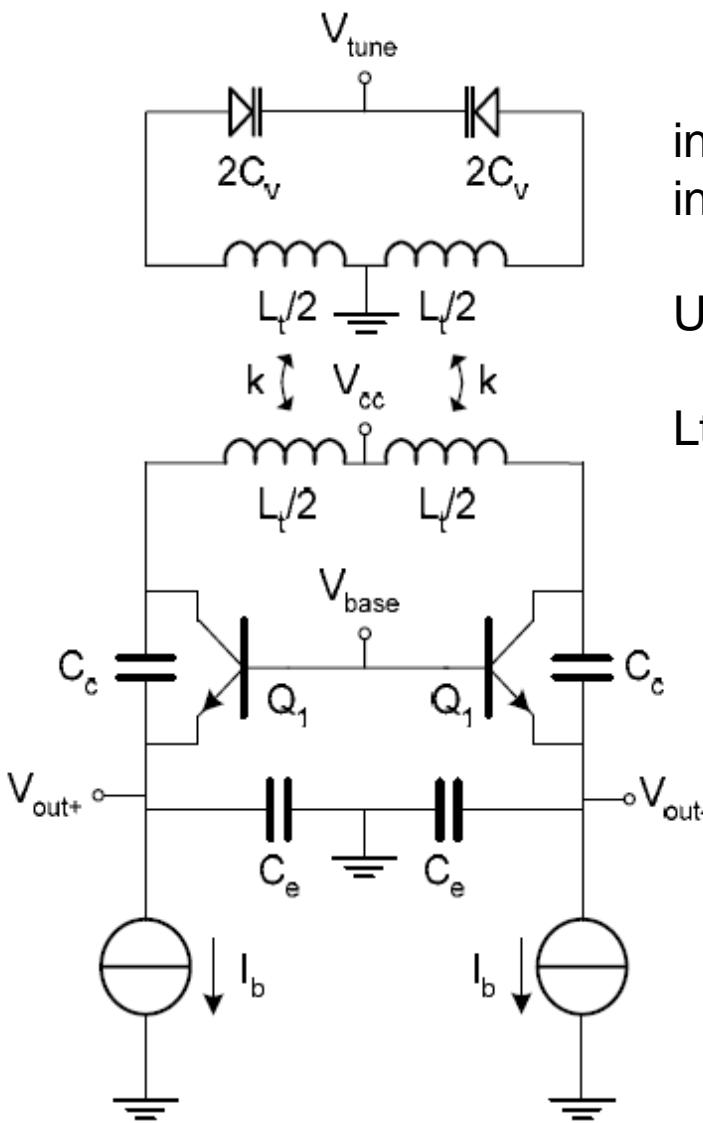
Measurements performed at 65°C

# Measurements: S11 vs. Gain

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# Some examples: Ka VCO



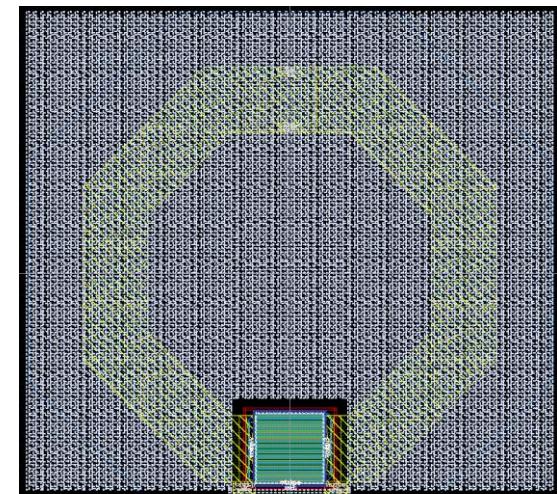
## Transformer-based Colpitts VCO

increase  $L \rightarrow$  increase in  $Q \rightarrow$  improves PN  
increase  $L \rightarrow$  increase tank impedance  $\rightarrow$  degrades PN

Use transformer:

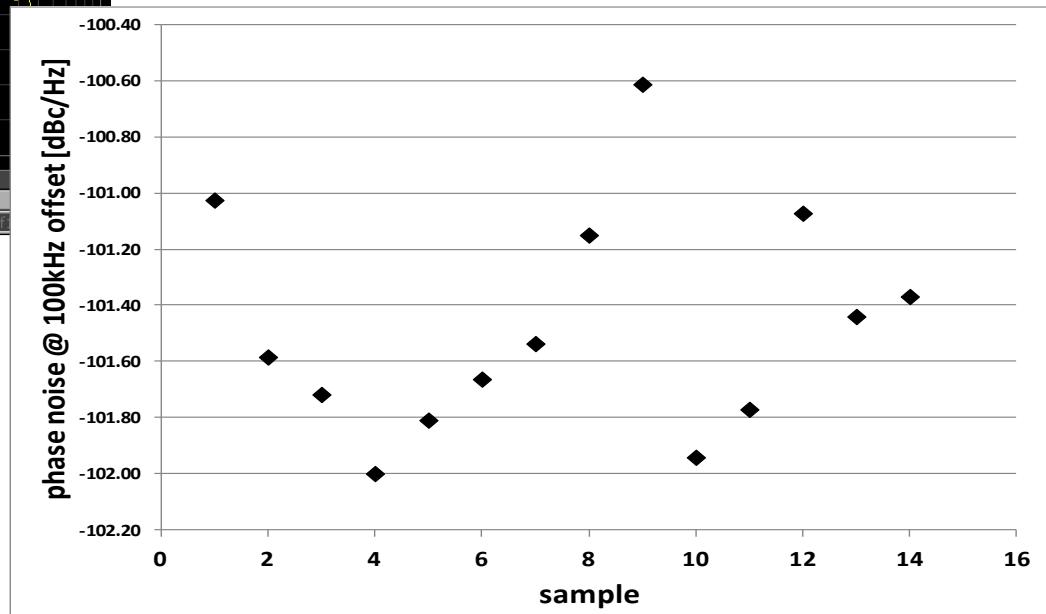
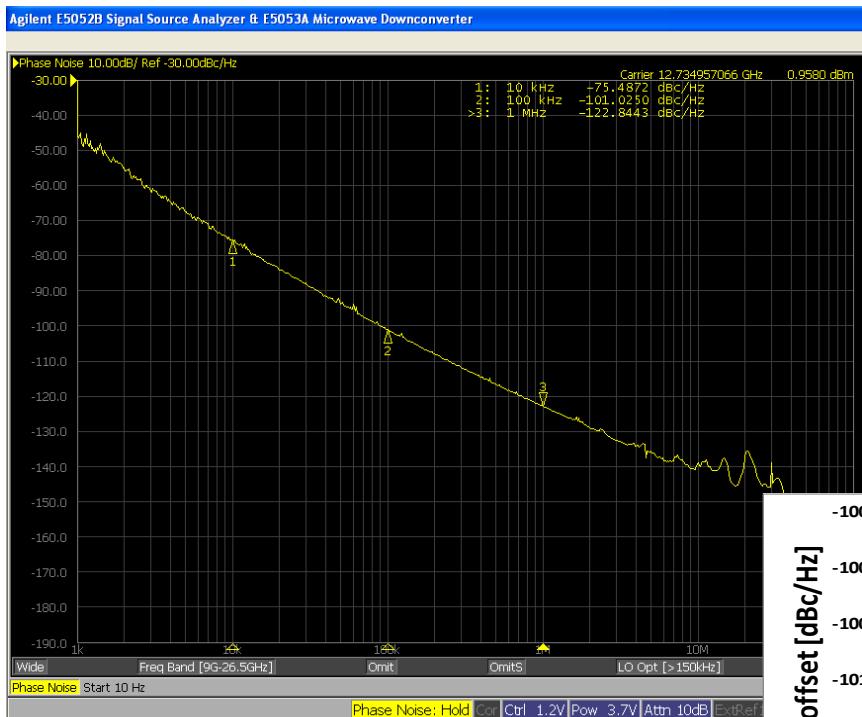
$L_{\text{transformer}} > L_{\text{tank}}$

Primary: M6  
Secondary: M5+M3

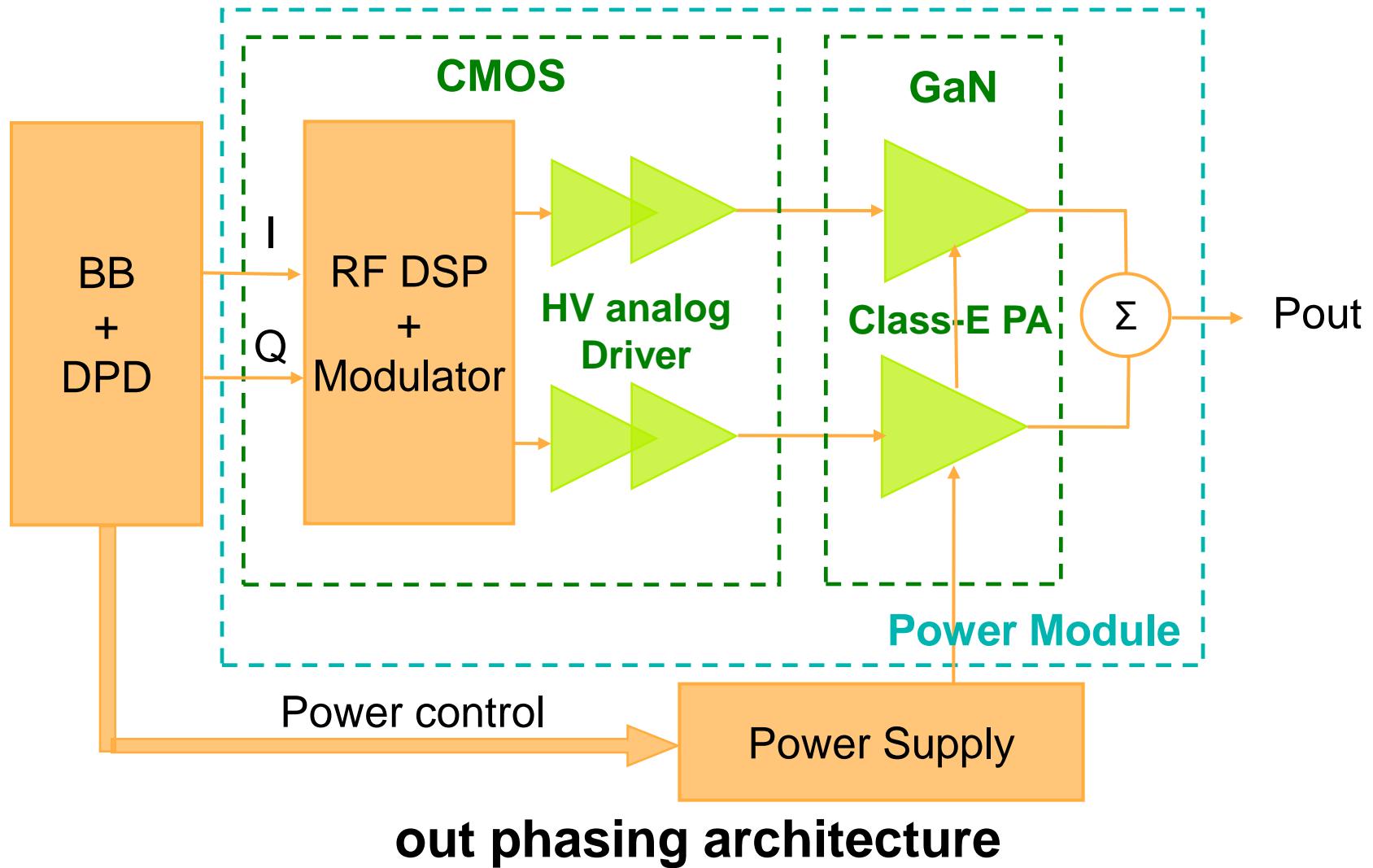


$Q=35 @ 17\text{GHz}$  for  $L=400\text{pH}$

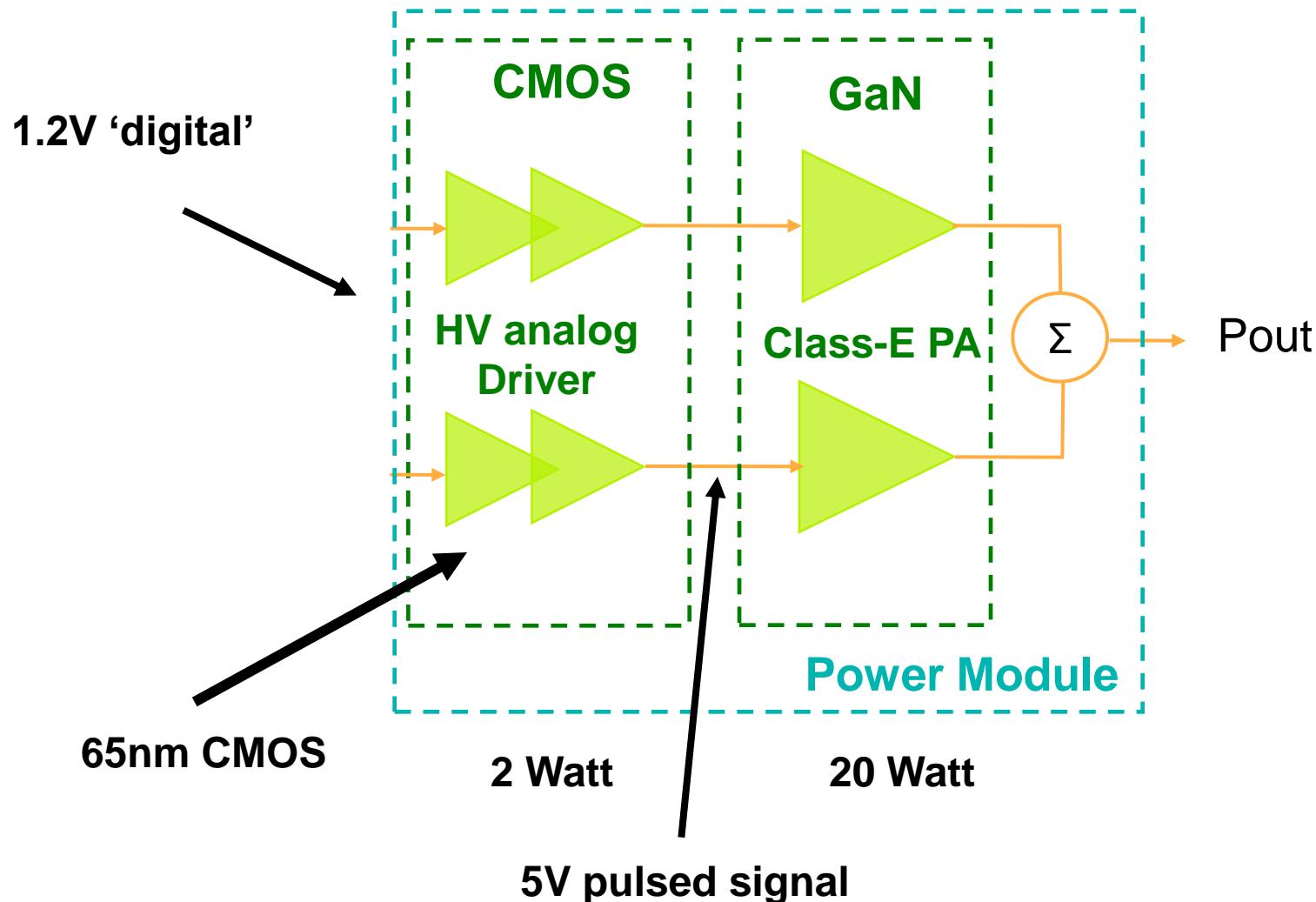
# Some examples: Ka VCO



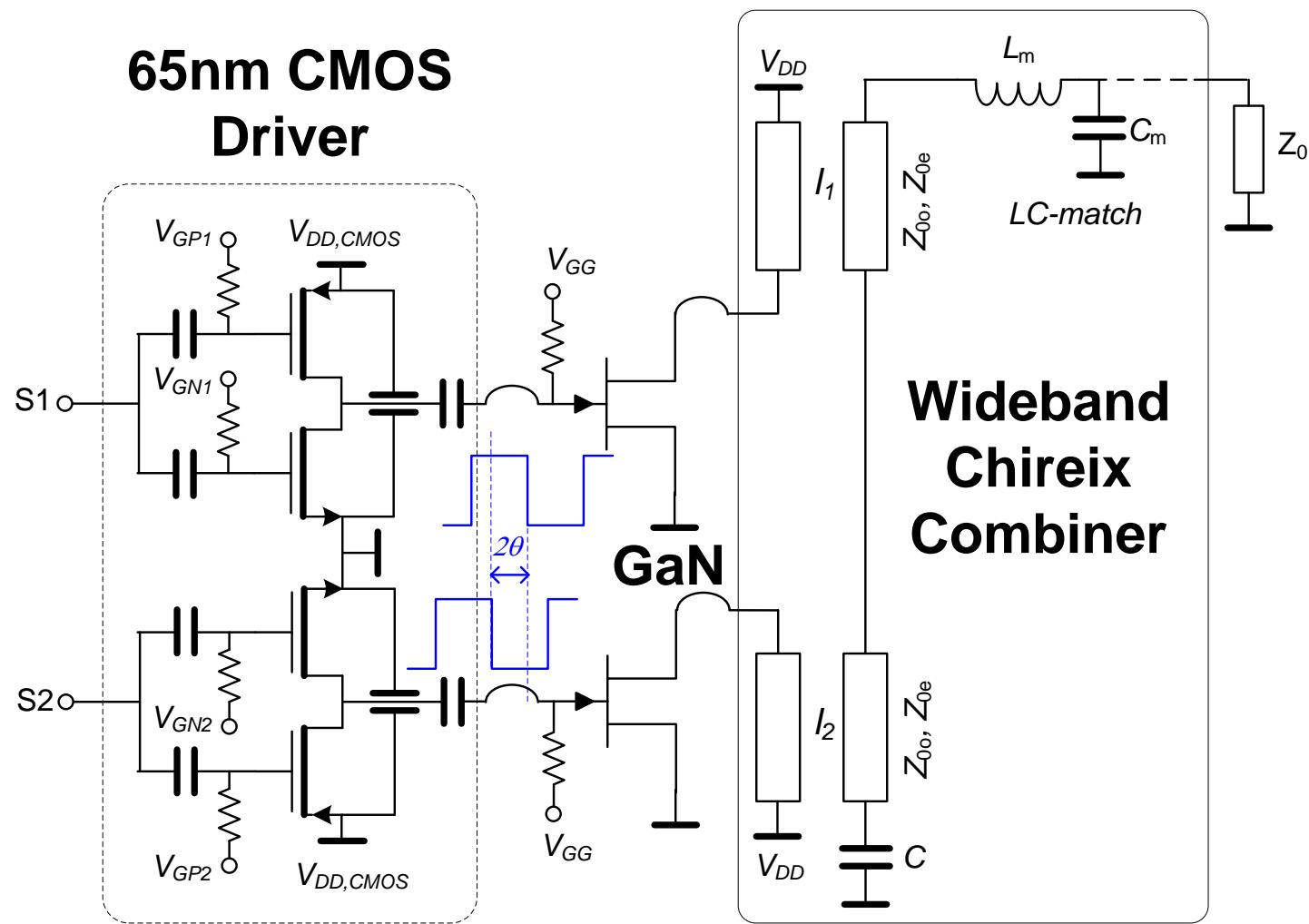
# Some examples: BS PA



# Some examples: BS PA

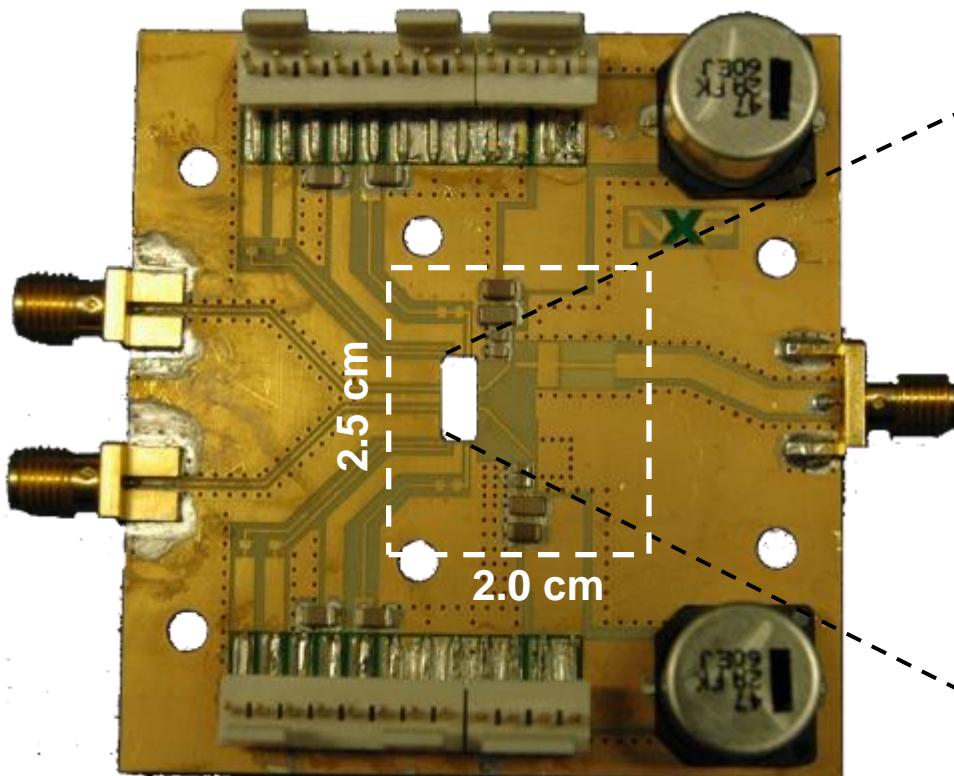


# Class-E Out-phasing PA

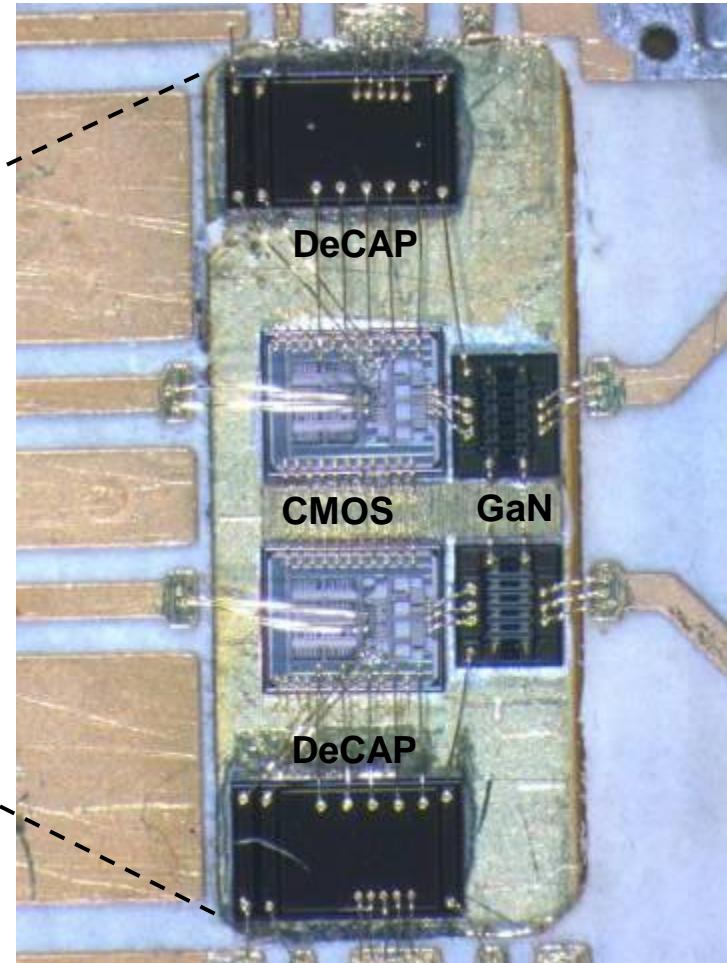


# Some examples: BS PA

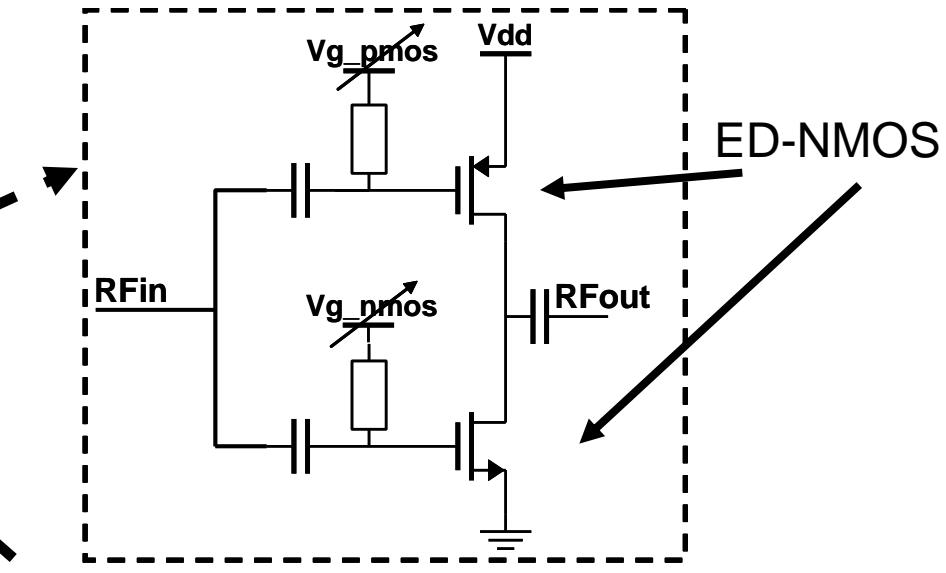
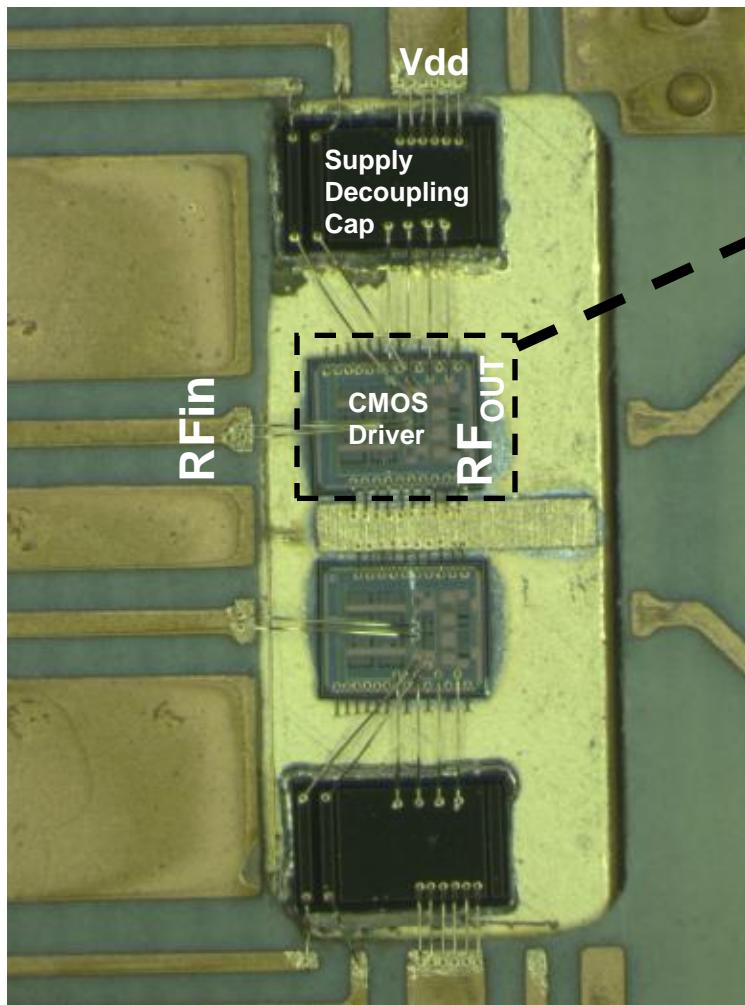
## 20W Out-phasing PA Line-up (CMOS + GaN)



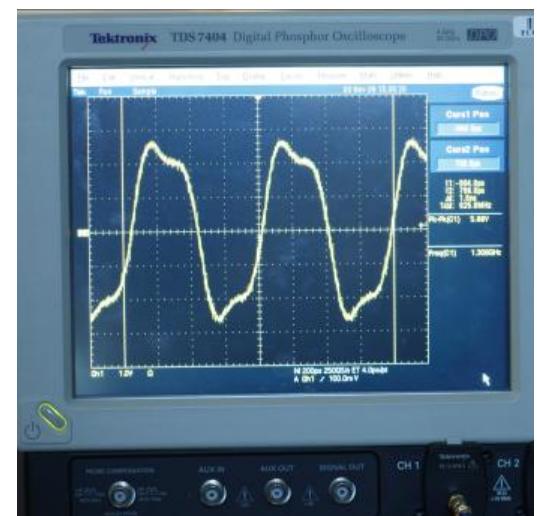
PA line up + matching: 5 cm<sup>2</sup>



# Some examples: BS PA

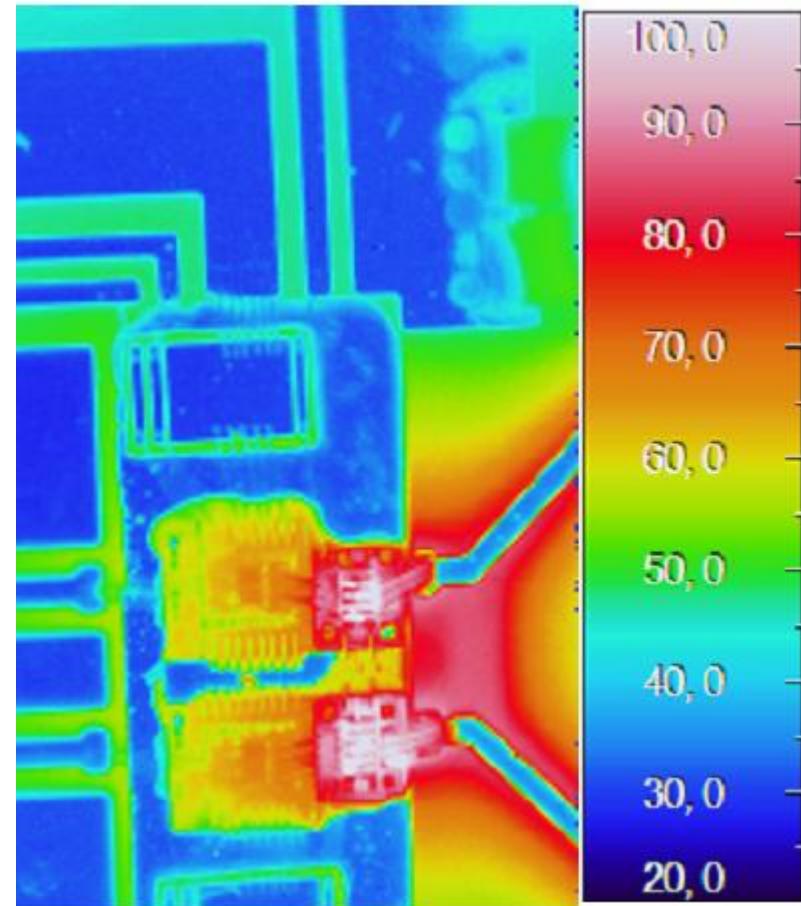
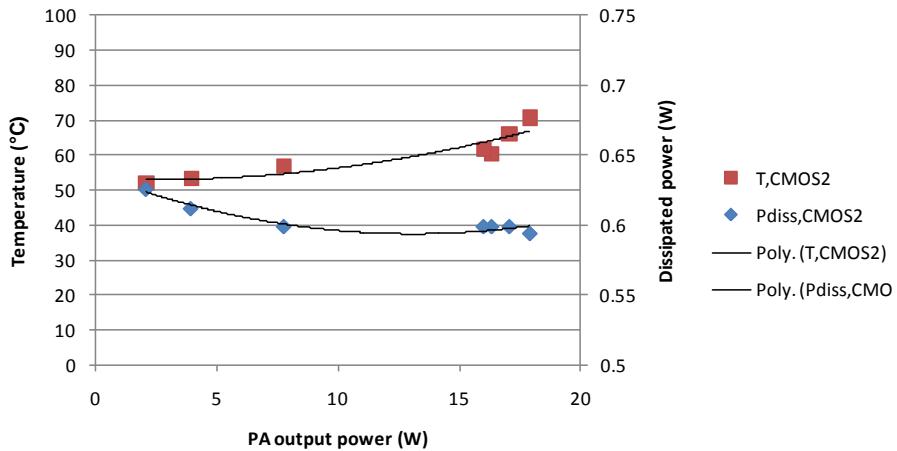


directly  $50\Omega$

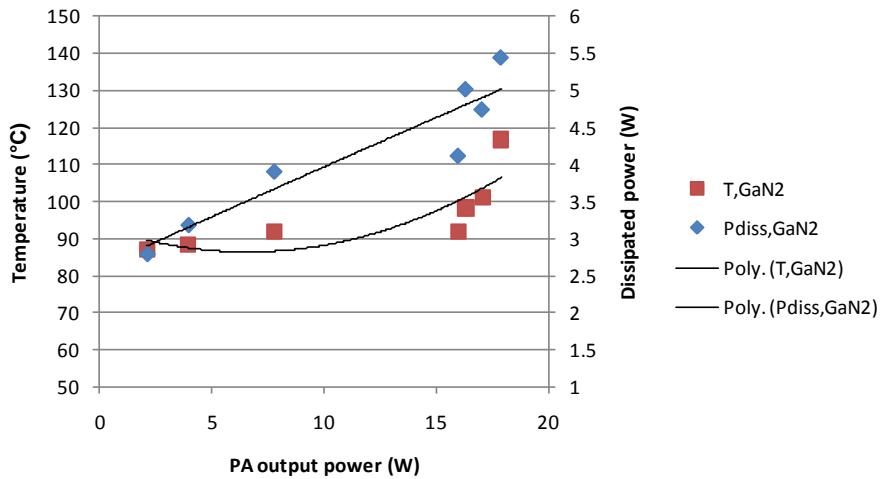


# Temperature gradients

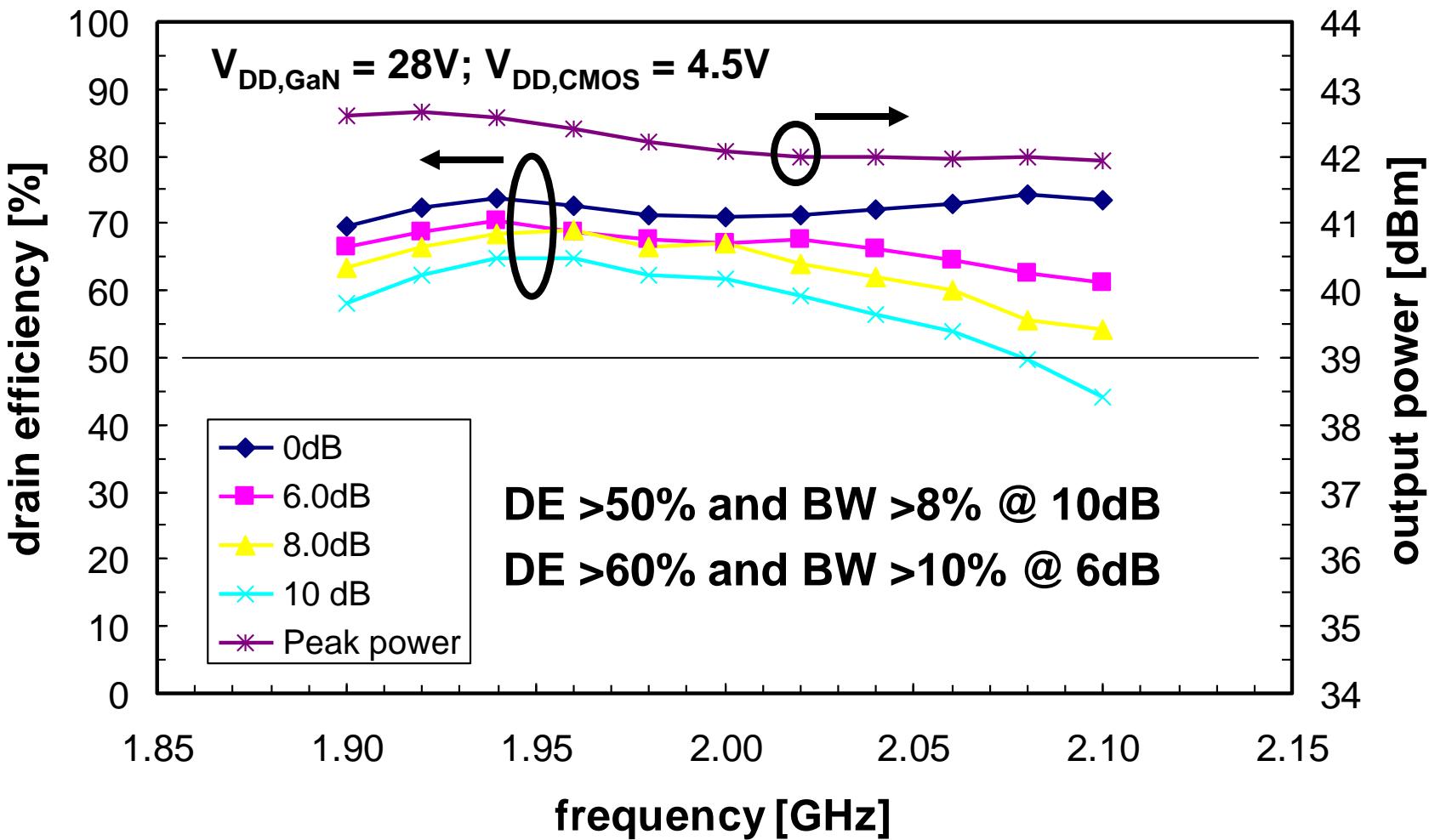
CMOS2 (top left)



GaN2 (top right)

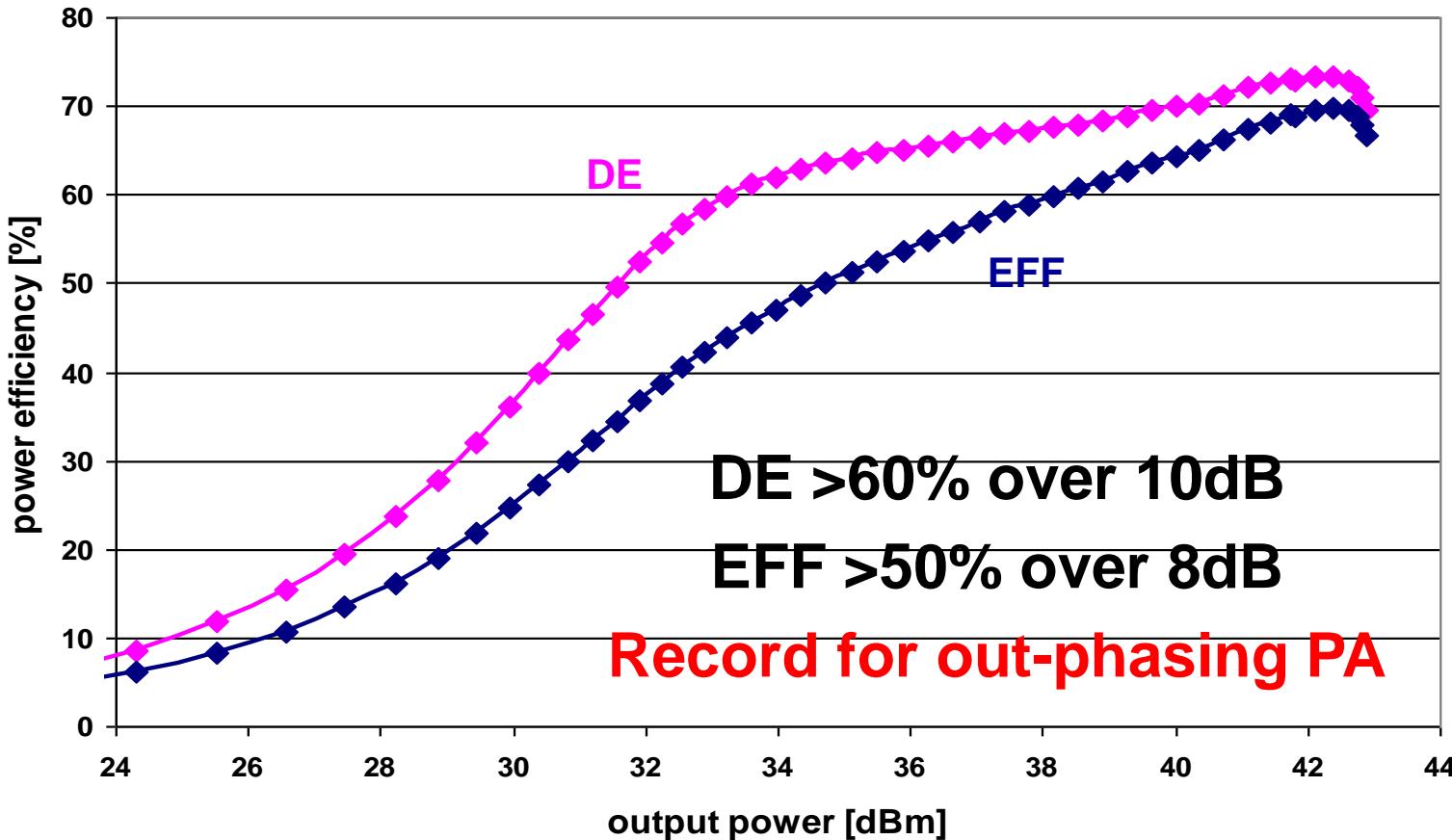


# Some examples: BS PA



# Some examples: BS PA

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# Concluding remarks

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- Wireless infrastructure IP dominated by compound technology, and for a reason!
- Modern Si-based technologies offer good RF performances, but still less than pHEMT technology
- Combination of improved architectures / circuits and Si-technology might lead to pHEMT replacement in wireless infrastructure

**Quest for High Performance RF**

# Acknowledgement

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- Team members RF ADT @ NXP
- N. Pulsford, N. Kramer, M. Geurts (NXP)